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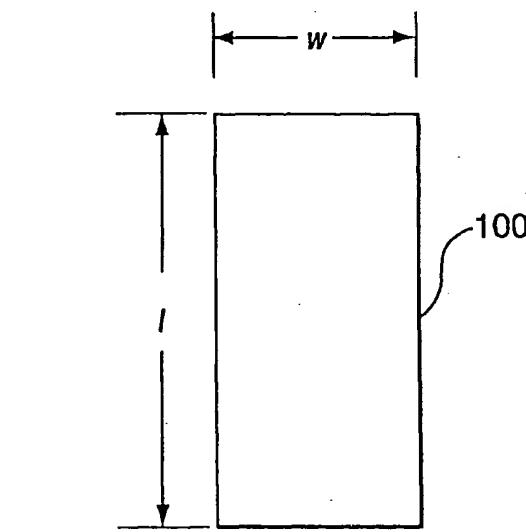
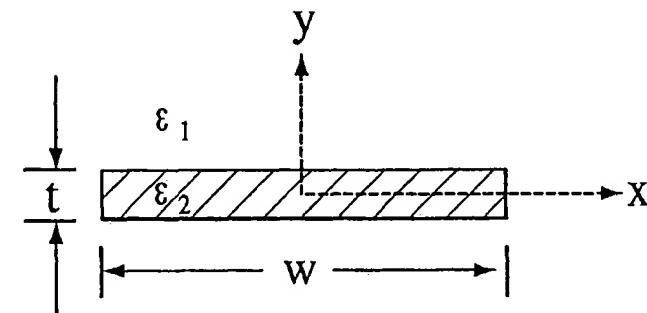
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(54) Title: OPTICAL WAVEGUIDE STRUCTURES



(57) Abstract: The purely bound electromagnetic modes of propagation (plasmon-polariton waves) supported by waveguide structures comprised of a thin lossy metal film of finite width embedded in an infinite homogeneous dielectric have been characterized at optical wavelengths. One of the fundamental modes supported by the structure evolves with decreasing film thickness and width towards the TEM wave supported by the background (an evolution similar to that exhibited by the Sb mode in symmetric metal film slab waveguides), its losses and phase constant tending asymptotically towards those of the TEM wave. Attenuation values can be well below those of the Sb mode supported by the corresponding metal film slab waveguide. Low mode power attenuation in the neighbourhood of 10 to 0.1 dB/cm is achievable at optical communication wavelengths, with even lower values being possible, thus enabling various devices to be constructed, such as couplers, splitters, modulators, interferometers, switches and periodic structures.

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OPTICAL WAVEGUIDE STRUCTURES

## TECHNICAL FIELD

The invention relates to optical devices and is especially applicable to waveguide structures and integrated optics.

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## BACKGROUND ART

This specification refers to several published articles. For convenience, the articles are cited in full in a numbered list at the end of the description and cited by number in the specification itself. The contents of these articles are incorporated herein 10 by reference and the reader is directed to them for reference.

In the context of this patent specification, the term "optical radiation" embraces electromagnetic waves having wavelengths in the infrared, visible and ultraviolet ranges.

The terms "finite" and "infinite" as used herein are used by persons skilled in this art to distinguish between waveguides having "finite" widths in which the actual width 15 is significant to the performance of the waveguide and the physics governing its operation and so-called "infinite" waveguides where the width is so great that it has no significant effect upon the performance and physics or operation.

At optical wavelengths, the electromagnetic properties of some metals closely resemble those of an electron gas, or equivalently of a cold plasma. Metals that 20 resemble an almost ideal plasma are commonly termed "noble metals" and include, among others, gold, silver and copper. Numerous experiments as well as classical electron theory both yield an equivalent negative dielectric constant for many metals when excited by an electromagnetic wave at or near optical wavelengths [1,2]. In a recent experimental study, the dielectric function of silver has been accurately measured 25 over the visible optical spectrum and a very close correlation between the measured dielectric function and that obtained via the electron gas model has been demonstrated [3].

It is well-known that the interface between semi-infinite materials having positive and negative dielectric constants can guide TM (Transverse Magnetic) surface waves.

30 In the case of a metal-dielectric interface at optical wavelengths, these waves are termed plasmon-polariton modes and propagate as electromagnetic fields coupled to surface plasmons (surface plasma oscillations) comprised of conduction electrons in the metal [4].

It is known to use a metal film of a certain thickness bounded by dielectrics above 35 and below as an optical slab (planar, infinitely wide) waveguiding structure, with the core of the waveguide being the metal film. When the film is thin enough, the plasmon-polariton modes guided by the interfaces become coupled due to field tunnelling through the metal, thus creating supermodes that exhibit dispersion with metal thickness. The

modes supported by infinitely wide symmetric and asymmetric metal film structures are well-known, as these structures have been studied by numerous researchers; some notable published works include references [4] to [10].

In general, only two purely bound TM modes, each having three field components, are guided by an infinitely wide metal film waveguide. In the plane perpendicular to the direction of wave propagation, the electric field of the modes is comprised of a single component, normal to the interfaces and having either a symmetric or asymmetric spatial distribution across the waveguide. Consequently, these modes are denoted  $s_b$  and  $a_b$  modes, respectively. The  $s_b$  mode can have a small attenuation constant and is often termed a long-range surface plasmon-polariton. The fields related to the  $a_b$  mode penetrate further into the metal than in the case of the  $s_b$  mode and can be much lossier by comparison. Interest in the modes supported by thin metal films has recently intensified due to their useful application in optical communications devices and components. Metal films are commonly employed in optical polarizing devices [11] while long-range surface plasmon-polaritons can be used for signal transmission [7]. In addition to purely bound modes, leaky modes are also known to be supported by these structures.

Infinitely wide metal film structures, however, are of limited practical interest since they offer one-dimensional (1-D) field confinement only, with confinement occurring along the vertical axis perpendicular to the direction of wave propagation, implying that modes will spread out laterally as they propagate from a point source used as the excitation. Metal films of finite width have recently been proposed in connection with polarizing devices [12], but merely as a cladding.

In addition to the lack of lateral confinement, plasmon-polariton waves guided by a metal-dielectric interface are in general quite lossy. Even long-range surface plasmons guided by a metal film can be lossy by comparison with dielectric waveguides. Known devices exploit this high loss associated with surface plasmons for the construction of plasmon-polariton based modulators and switches. Generally, known plasmon-polariton based modulator and switch devices can be classified along two distinct architectures. The first architecture is based on the phenomenon of attenuated total reflection (ATR) and the second architecture is based on mode coupling between a dielectric waveguide and a nearby metal. Both architectures depend on the dissipation of optical power within an interacting metal structure.

ATR based devices depend on the coupling of an optical beam, which is incident upon a dielectric-metal structure placed in optical proximity, to a surface plasmon-polariton mode supported by the metal structure. At a specific angle of incidence, which depends on the materials used and the particular geometry of the device, coupling to a plasmon mode is maximised and a drop in the power reflected from the metal surface

is observed. ATR based modulators make use of this attenuated reflection phenomenon along with means for varying electrically or otherwise at least one of the optical parameters of one of the dielectrics bounding the metal structure in order to shift the angle of incidence where maximum coupling to plasmons occurs. Electrically shifting the  
5 angle of maximum coupling results in a modulation of the intensity of the reflected light. Examples of devices that are based on this architecture are disclosed in references [23] to [36].

Mode coupling devices are based on the optical coupling of light propagating in a dielectric waveguide to a nearby metal film placed a certain distance away and in  
10 parallel with the dielectric waveguide. The coupling coefficient between the optical mode propagating in the waveguide and the plasmon-polariton mode supported by the nearby metal film is adjusted via the materials selected and the geometrical parameters of the device. Means is provided for varying, electrically or otherwise, at least one of the optical parameters of one of the dielectrics bounding the metal. Varying an optical  
15 parameter (the index of refraction, say) varies the coupling coefficient between the optical wave propagating in the dielectric waveguide and the lossy plasmon-polariton wave supported by the metal. This results in a modulation in the intensity of the light exiting the dielectric waveguide. References [37] to [40] disclose various device implementations based upon this phenomenon. Reference [41] further discusses the  
20 physical phenomenon underlying the operation of these devices.

Reference [42] discusses an application of the ATR phenomenon for realising an optical switch or bistable device.

These known modulation and switching devices disadvantageously require relative high control voltages and have limited electrical/optical bandwidth.

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#### DISCLOSURE OF INVENTION:

The present invention seeks to eliminate, or at least mitigate, one or more of the disadvantages of the prior art.

According to one aspect of the present invention there is provided a waveguide  
30 structure comprising a thin strip having finite width and thickness with dimensions such that optical radiation having a wavelength in a predetermined range couples to the strip and propagates along the length of the strip as a plasmon-polariton wave. The strip may comprise a material having a relatively high free charge carrier density, for example a conductor or certain classes of highly-doped semiconductor. The surrounding material  
35 may have a relatively low free charge carrier density, i.e. an insulator or undoped lightly doped semiconductor.

Such a strip of finite width offers two-dimensional (2-D) confinement in the transverse plane, i.e. perpendicular to the direction of propagation, and, since suitable

low-loss waveguides can be fabricated from such strip, it may be useful for signal transmission and routing or to construct components such as couplers, power splitters, interferometers, modulators, switches and other typical components of integrated optics. In such devices, different sections of the strip serving different functions, in some cases 5 in combination with additional electrodes. The strip sections may be discrete and concatenated or otherwise interrelated, or sections of one or more continuous strips.

For example, where the optical radiation has a free-space wavelength of 1550 nm, and the waveguide is made of a strip of a noble metal surrounded by a good dielectric, say glass, suitable dimensions for the strip are thickness less than about 0.1 microns, 10 preferably about 20 nm, and width of a few microns, preferably about 4 microns.

The strip could be straight, curved, bent, tapered, and so on.

The dielectric material may be inhomogeneous, for example a combination of slabs, strips, laminae, and so on. The conductive or semiconductive strip may be inhomogeneous, for example a gold layer sandwiched between thin layers of titanium.

15 The plasmon-polariton wave which propagates along the structure may be excited by an appropriate optical field incident at one of the ends of the waveguide, as in an end-fire configuration, and/or by a different radiation coupling means.

The low free-carrier density material may comprise two distinct portions with the strip extending therebetween, at least one of the two distinct portions having at least one 20 variable electromagnetic property, and the device then may further comprise means for varying the value of said electromagnetic property of said one of the portions so as to vary the propagation characteristics of the waveguide structure and the propagation of the plasmon-polariton wave.

In some embodiments of the invention, for one said value of the electromagnetic 25 property, propagation of the plasmon-polariton wave is supported and, for another value of said electromagnetic property, propagation of the plasmon-polariton wave is at least inhibited. Such embodiments may comprise modulators or switches.

Different embodiments of the invention may employ different means of varying the electromagnetic property, such as varying the size of at least one of said portions, 30 especially if it comprises a fluid.

The at least one variable electromagnetic property of the material may comprise permittivity, permeability or conductivity.

Where the portion comprises an electro-optic material, the variable electromagnetic property will be permittivity, which may be varied by applying an 35 electric field to the portion, or changing an electric field applied thereto, using suitable means.

Where the portion comprises a magneto-optic material, the variable electromagnetic property will be permittivity which may be varied by applying a

magnetic field to the portion or changing a magnetic field applied thereto, using suitable means.

Where the portion comprises a thermo-optic material, the electromagnetic property may be permittivity and be varied by changing the temperature of the material.

5 Where the portion comprises an acousto-optical (photoelastic) material, the electromagnetic property may be permittivity and be varied by changing mechanical strain in the material.

Where the portion comprises a magnetic material (such as a ferrite), the electromagnetic property will be permeability and may be varied by applying a magnetic 10 field to the material or changing a magnetic field applied thereto, by suitable means.

Where the portion comprises a semiconductor material, the electromagnetic property will be conductivity or permittivity and may be varied by changing free charge carrier density in said portion, using suitable means.

Additionally or alternatively, the propagation of the plasmon-polariton wave may 15 be varied by varying an electromagnetic property of the strip. For example, the permittivity of the strip may be varied by changing the free charge carrier density or by changing or applying a magnetic field through the strip.

Various objects, features, aspects and advantages of the present invention will become more apparent from the following detailed description, taken in conjunction with 20 the accompanying drawings, of a preferred embodiment of the invention.

#### BRIEF DESCRIPTION OF THE DRAWINGS

Figures 1(a) and 1(b) are a cross-sectional illustration and a plan view, respectively, of a symmetric waveguide structure embodying the present invention in 25 which the core is comprised of a lossy metal film of thickness  $t$ , width  $w$ , length  $l$  and permittivity  $\epsilon_2$  embedded in a cladding or background comprising an "infinite" homogeneous dielectric having a permittivity  $\epsilon_1$ ;

Figures 2(a) and 2(b) illustrate dispersion characteristics with thickness of the first eight modes supported by a symmetric metal film waveguide of width  $w = 1\mu\text{m}$ . The 30  $a_b$  and  $s_b$  modes supported for the case  $w = \infty$  are shown for comparison. (a) Normalized phase constant; (b) Normalized attenuation constant;

Figures 3(a),(b),(c),(d),(e) and (f) illustrate the spatial distribution of the six field components related to the  $ss_b^0$  mode supported by a symmetric metal film waveguide of thickness  $t = 100\text{nm}$  and width  $w = 1\mu\text{m}$ . The waveguide cross-section is located 35 in the  $x - y$  plane and the metal is bounded by the region  $-0.5 \leq x \leq 0.5\mu\text{m}$  and  $-0.05 \leq y \leq 0.05\mu\text{m}$ ; outlined as the rectangular dashed contour. The field distributions are normalized such that  $\max|Re\{E_y\}| = 1$ ;

Figures 4(a),(b),(c),(d),(e) and (f) illustrate the spatial distribution of the six field components related to the  $sa_b^0$  mode supported by a symmetric metal film waveguide of thickness  $t = 100nm$  and width  $w = 1\mu m$ . The waveguide cross-section is located in the  $x - y$  plane and the metal is bounded by the region  $-0.5 \leq x \leq 0.5\mu m$  and  $-0.05 \leq y \leq 0.05\mu m$ , outlined as the rectangular dashed contour. The field distributions are normalized such that  $\max |Re\{E_y\}| = 1$ ;

Figures 5(a),(b),(c),(d),(e) and (f) illustrate the spatial distribution of the six field components related to the  $as_b^0$  mode supported by a symmetric metal film waveguide of thickness  $t = 100nm$  and width  $w = 1\mu m$ . The waveguide cross-section is located in the  $x - y$  plane and the metal is bounded by the region  $-0.5 \leq x \leq 0.5\mu m$  and  $-0.05 \leq y \leq 0.05\mu m$ , outlined as the rectangular dashed contour. The field distributions are normalized such that  $\max |Re\{E_y\}| = 1$ ;

Figures 6(a),(b),(c),(d),(e) and (f) illustrate the spatial distribution of the six field components related to the  $aa_b^0$  mode supported by a symmetric metal film waveguide of thickness  $t = 100nm$  and width  $w = 1\mu m$ . The waveguide cross-section is located in the  $x - y$  plane and the metal is bounded by the region  $-0.5 \leq x \leq 0.5\mu m$  and  $-0.05 \leq y \leq 0.05\mu m$ , outlined as the rectangular dashed contour. The field distributions are normalized such that  $\max |Re\{E_y\}| = 1$ ;

Figures 7(a),(b),(c),(d),(e) and (f) are contour plots of  $Re\{S_z\}$  associated with the  $ss_b^0$  mode for symmetric metal film waveguides of width  $w = 1\mu m$  and various thicknesses. The power confinement factor  $cf$  is also given in all cases, and is computed via equation (12) with the area of the waveguide core  $A_c$  taken as the area of the metal region. In all cases, the outline of the metal film is shown as the rectangular dashed contour;

Figure 8 illustrates a normalized profile of  $Re\{S_z\}$  associated with the  $ss_b^0$  mode for a symmetric metal film waveguide of width  $w = 1\mu m$  and thickness  $t = 20nm$ . The waveguide cross-section is located in the  $x - y$  plane and the metal film is bounded by the region  $-0.5 \leq x \leq 0.5\mu m$  and  $-0.01 \leq y \leq 0.01\mu m$ , outlined as the rectangular dashed contour;

Figures 9(a),(b),(c) and (d) illustrate the spatial distribution of the  $E_y$  field component related to some higher order modes supported by a symmetric metal film waveguide of thickness  $t = 100nm$  and width  $w = 1\mu m$ . In all cases, the waveguide cross-section is located in the  $x - y$  plane and the metal film is bounded by the region  $-0.5 \leq x \leq 0.5\mu m$  and  $-0.05 \leq y \leq 0.05\mu m$ , outlined as the rectangular dashed contour;

Figures 10(a) and (b) illustrate dispersion characteristics with thickness of the first six modes supported by a symmetric metal film waveguide of width  $w = 0.5\mu m$ . The

$a_b$  and  $s_b$  modes supported for the case  $w = \infty$  are shown for comparison. (a) Normalized phase constant; (b) Normalized attenuation constant;

Figures 11(a) and (b) illustrate dispersion characteristics with thickness of the  $ss_b^0$  mode supported by symmetric metal film waveguides of various widths. The 5  $s_b$  mode supported for the case  $w = \infty$  is shown for comparison. (a) Normalized phase constant; (b) Normalized attenuation constant;

Figures 12(a),(b),(c) and (d) illustrate a contour plot of  $Re\{S_z\}$  associated with the  $ss_b^0$  mode for symmetric metal film waveguides of thickness  $t = 20nm$  and various widths. The power confinement factor  $\alpha$  is also given in all cases, and is computed via 10 equation (12) with the area of the waveguide core  $A_c$  taken as the area of the metal region. In all cases, the outline of the metal film is shown as the rectangular dashed contour;

Figure 13 illustrates dispersion characteristics with thickness of the  $ss_b^0$  mode supported by a symmetric metal film waveguide of width  $w = 0.5\mu m$  for various 15 background permittivities  $\epsilon_{r,1}$ . The normalized phase constant is plotted on the left axis and the normalized attenuation constant is plotted on the right one;

Figures 14(a),(b),(c) and (d) illustrates a contour plot of  $Re\{S_z\}$  associated with the  $ss_b^0$  mode for a symmetric metal film waveguide of width  $w = 0.5\mu m$  and thickness  $t = 20nm$  for various background permittivities  $\epsilon_{r,1}$ . In all cases, the outline of the metal 20 film is shown as the rectangular dashed contour;

Figures 15(a) and (b) illustrate dispersion characteristics with frequency of the  $ss_b^0$  mode supported by symmetric metal film waveguides of width  $w = 0.5\mu m$  and  $w = 1\mu m$  and various thicknesses  $t$ . The  $s_b$  mode supported for the case  $w = \infty$  and the thicknesses considered is shown for comparison. (a) Normalized phase constant. (b) 25 Mode power attenuation computed using Equation (16) and scaled to  $dB/cm$ ;

Figures 16(a),(b),(c),(d),(e) and (f) illustrate a contour plot of  $Re\{S_z\}$  associated with the  $ss_b^0$  mode for symmetric metal film waveguides of width  $w = 0.5\mu m$  and  $w = 1\mu m$ , and thickness  $t = 20nm$  at various free-space wavelengths of excitation  $\lambda_0$ . In all cases, the outline of the metal film is shown as the rectangular dashed contour;

30 Figures 17(a) and 17(b) are a cross-sectional view and a plan view, respectively, of a second embodiment of the invention in the form of an asymmetric waveguide structure formed by a core comprising a lossy metal film of thickness  $t$ , width  $w$  and permittivity  $\epsilon_2$  supported by a homogeneous semi-infinite substrate of permittivity  $\epsilon_1$  and with a cover or superstrate comprising a homogeneous semi-infinite dielectric of 35 permittivity  $\epsilon_3$ ;

Figures 18(a) and 18(b) illustrate dispersion characteristics with thickness of the first seven modes supported by an asymmetric metal film waveguide of width  $w = 1\mu m$ .

The  $a_b$  and  $s_b$  modes supported for the case  $w = \infty$  are shown for comparison. (a) Normalized phase constant. (b) Normalized attenuation constant;

Figures 19(a),(b),(c) and (d) illustrate spatial distribution of the  $E_y$  field component related to the  $ss_b^0$  mode supported by an asymmetric metal film waveguide 5 of width  $w = 1 \mu\text{m}$  for four film thicknesses. The waveguide cross-section is located in the x-y plane and the metal region is outlined as the rectangular dashed contour. The field distributions are normalized such that  $\max |\Re\{E_y\}| = 1$ ;

Figures 20(a),(b),(c) and (d) illustrate spatial distribution of the  $E_y$  field component related to two higher order modes supported by an asymmetric metal film 10 waveguide of width  $w = 1 \mu\text{m}$  for two film thicknesses. In all cases, the waveguide cross-section is located in the x-y plane and the metal region is outlined as the rectangular dashed contour. The field distributions are normalized such that  $\max |\Re\{E_y\}| = 1$ ;

Figures 21(a) and (b) illustrate dispersion characteristics with thickness of the first 15 six modes supported by an asymmetric metal film waveguide of width  $w = 1 \mu\text{m}$ . The  $a_b$  and  $s_b$  modes supported for the case  $w = \infty$  are shown for comparison. (a) Normalized phase constant. (b) Normalized attenuation constant;

Figures 22(a),(b),(c) and (d) illustrate spatial distribution of the  $E_y$  field component related to modes supported by an asymmetric metal film waveguide of width 20  $w = 1 \mu\text{m}$ . In all cases, the waveguide cross-section is located in the x-y plane and the metal region is outlined as the rectangular dashed contour. The field distributions are normalized such that  $\max |\Re\{E_y\}| = 1$ ;

Figures 23(a) and 23(b) illustrate dispersion characteristics with thickness of the first six modes supported by an asymmetric metal film waveguide of width  $w = 0.5 \mu\text{m}$ . 25 The  $a_b$  and  $s_b$  modes supported for the case  $w = \infty$  are shown for comparison. (a) Normalized phase constant. (b) Normalized attenuation constant;

Figures 24(a) and 24(b) illustrate dispersion characteristics with thickness of the  $ss_b^0$  and  $sa_b^1$  modes supported by an asymmetric metal film waveguide of width  $w = 0.5 \mu\text{m}$  for various cases of  $\epsilon_3$ . (a) Normalized phase constant; the inset shows an 30 enlarged view of the region bounded by  $0.04 \leq t \leq 0.08 \mu\text{m}$  and  $2.0 \leq \beta/\beta_0 \leq 2.3$ . (b) Normalized attenuation constant; the inset shows an enlarged view of the region bounded by  $0.05 \leq t \leq 0.08 \mu\text{m}$  and  $7.0 \times 10^{-3} \leq \alpha/\beta_0 \leq 2.0 \times 10^{-2}$ ;

Figures 25(a),(b),(c) and (d) illustrate spatial distribution of the  $E_y$  field component related to the  $sa_b^1$  mode supported by an asymmetric metal film waveguide 35 of width  $w = 0.5 \mu\text{m}$  for four film thicknesses. The waveguide cross-section is located in the x-y plane and the metal region is outlined as the rectangular dashed contour. The field distributions are normalized such that  $\max |\Re\{E_y\}| = 1$ ;

Figures 26(a),(b),(c) and (d) illustrate a contour plot of  $\Re\{S_2\}$  associated with the long-ranging modes supported by asymmetric metal film waveguides of width  $w = 0.5 \mu\text{m}$  and having different superstrate permittivities  $\epsilon_3$ . In all cases, the outline of the metal film is shown as the rectangular dashed contour;

5 Figure 27 is a plan view of a waveguide with opposite sides stepped to provide different widths;

Figure 28 is a plan view of a waveguide which is tapered and slanted;

Figure 29 is a plan view of a trapezoidal waveguide;

10 Figure 30 is a plan view of a waveguide having curved side edges and suitable for use as a transition piece;

Figure 31 is a plan view of a curved waveguide section suitable for interconnecting waveguides at a corner;

Figure 32 is a plan view of a two-way splitter/combiner formed by a combination of three straight waveguide sections and one tapered waveguide section;

15 Figure 33 is a plan view of an angled junction using a slanted section;

Figure 34 is a plan view of a power divider formed by a trapezoidal section and pairs of concatenated bends;

Figure 35 is a plan view of a Mach-Zehnder interferometer formed using a combination of the waveguide sections;

20 Figure 36(a) is a schematic plan view of a modulator using the Mach-Zehnder waveguide structure of Figure 35;

Figures 36(b) and 36(c) are inset diagrams illustrating alternative ways of applying a modulation control voltage;

Figure 37 is a plan view of a modulator using the Mach-Zehnder waveguide structure of Figure 35 and illustrating magnetic field control;

Figure 38 is a plan view of a periodic structure formed by a series of unit cells each comprising two waveguide sections having different widths and lengths;

Figure 39 is a plan view of a periodic waveguide structure formed by a series of unit cells each comprising two opposed trapezoidal waveguide sections;

30 Figure 40(a) is a plan view of an edge coupler formed by two parallel strips of straight waveguide with various other waveguides for coupling signals to and from them;

Figure 40(b) is an inset diagram illustrating a way of applying a modulation control voltage;

Figure 41(a) is a perspective view of an edge coupler in which the parallel strips are not co-planar;

Figure 41(b) is an inset diagram illustrating a way of applying a modulation control voltage;

Figure 42 is a plan view of an intersection formed by four sections of waveguide;

Figures 43(a) and 43(b) are a schematic front view and corresponding top plan view of an electro-optic modulator employing the waveguide structure of Figure 17(a);

Figures 44(a) and 44(b) are a schematic front view and corresponding top view of an alternative electro-optic modulator also using the waveguide structure of Figure 5 17(a);

Figure 44(c) illustrates an alternative connection arrangement of the modulator of Figure 44(a);

Figure 45 is a schematic front view of a third embodiment of electro-optic modulator also using the waveguide structure of Figure 17(a);

10 Figure 46 is a schematic front view of a magneto-optic modulator also using the waveguide structure of Figure 17(a);

Figure 47 is a schematic front view of a thermo-optic modulator also using the waveguide structure of Figure 17(a);

15 Figure 48 is a schematic perspective view of an electro-optic switch also using the waveguide structure of Figure 17(a);

Figure 49 is a schematic perspective view of a magneto-optic switch also using the waveguide structure of Figure 17(a);

Figure 50 is a schematic perspective view of a thermo-optic switch also using the waveguide structure of Figure 17(a);

20 Figure 51 gives the mode power attenuation for metal film waveguides of various widths and thicknesses. The metal used is Au and the background dielectric is SiO<sub>2</sub>. The optical free-space wavelength of analysis is set to  $\lambda_0 = 1.55 \mu\text{m}$ ; and

Figure 52 gives the mode power attenuation for metal film waveguides of various widths and thicknesses. The metal used is Al and the background dielectric is SiO<sub>2</sub>. The 25 optical free-space wavelength of analysis is set to  $\lambda_0 = 1.55 \mu\text{m}$ .

## DESCRIPTION OF THE PREFERRED EMBODIMENTS

### I. Introduction

In order to facilitate an understanding of the specific optical devices embodying 30 the invention, their theoretical basis will first be explained with reference to Figures 1 to 26(d).

The following is a comprehensive description of the purely bound modes of propagation supported by symmetric and asymmetric waveguide structures comprised of a thin lossy metal film of *finite-width* as the core. The core can be embedded in an 35 "infinite" homogeneous dielectric medium as shown in Figure 1(a) or supported by a semi-infinite homogeneous dielectric substrate and covered by a different semi-infinite homogeneous dielectric superstrate as shown in Figure 17(a). The description is organized as follows. Section II summarizes the physical basis and numerical technique

used to analyze the structures of interest. Sections III through VI describe the modes supported by symmetric structures as shown in Figure 1(a) and sections VII through X describe the modes supported by asymmetric structures as shown in Figure 17(a). Concluding remarks are given in section XI.

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## II. Physical Basis and Numerical Technique

A symmetric structure embodying the present invention is shown in Figures 1(a) and 1(b). It comprises a lossy metal film of thickness  $t$ , width  $w$  and equivalent permittivity  $\epsilon_2$ , surrounded by a cladding or background comprising an infinite homogeneous dielectric of permittivity  $\epsilon_1$ . Figure 17(a) shows an asymmetric structure ( $\epsilon_1 \neq \epsilon_3$ ) embodying the present invention. The Cartesian coordinate axes used for the analysis are also shown with propagation taking place along the  $z$  axis, which is out of the page.

It is assumed that the metal region shown in Figures 1(a) and 17(a) can be modeled as an electron gas over the wavelengths of interest. According to classical or Drude electron theory, the complex relative permittivity of the metal region is given by the well-known plasma frequency dispersion relation [4]:

20

$$\epsilon_{r,2} = \left( 1 - \frac{\omega_p^2}{\omega^2 + \nu^2} \right) - j \left( \frac{\omega_p^2 \nu}{\omega(\omega^2 + \nu^2)} \right) \quad (1)$$

25

where  $\omega$  is the excitation frequency,  $\omega_p$  is the electron plasma frequency and  $\nu$  is the effective electron collision frequency, often expressed as  $\nu = 1/\tau$  with  $\tau$  defined as the relaxation time of electrons in the metal. When  $\omega^2 + \nu^2 < \omega_p^2$  (which is the case for many metals at optical wavelengths) a negative value for the real part  $\epsilon_{r,2}$  is obtained, implying that plasmon-polariton modes can be supported at interfaces with normal dielectrics.

### Electromagnetic Wave and Field Equations

The modes supported by the structures are obtained by solving a suitably defined boundary value problem based on Maxwell's equations written in the frequency domain for a lossy inhomogeneous isotropic medium. Uncoupling Maxwell's equations yields the following time-harmonic vectorial wave equations for the  $\mathbf{E}$  and  $\mathbf{H}$  fields:

$$\nabla \times \nabla \times \mathbf{E} - \omega^2 \epsilon(x,y) \mu \mathbf{E} = 0 \quad (2)$$

$$\nabla \times \epsilon(x,y)^{-1} \nabla \times \mathbf{H} - \omega^2 \mu \mathbf{H} = 0 \quad (3)$$

35 where the permittivity  $\epsilon$  is a complex function of cross-sectional space, and describes the waveguide structure. For the structures analyzed in this description,  $\mu$  is homogeneous and taken as the permeability of free space  $\mu_0$ .

Due to the nature of the numerical method used to solve the boundary value problem, the implicit  $y$  dependence of the permittivity can be immediately removed since any inhomogeneity along  $y$  is treated by dividing the structure into a number of layers that are homogeneous along this direction, and suitable boundary conditions are applied 5 between them.

The two vectorial wave equations (2) and (3) are expanded in each layer into scalar wave equations, some being coupled by virtue of the remaining inhomogeneity in  $\epsilon$  along  $x$ . Since the structure under consideration is invariant along the propagation axis (taken to be in the  $+z$  direction), the mode fields vary along this dimension according 10 to  $e^{-\gamma z}$  where  $\gamma = \alpha + j\beta$  is the complex propagation constant of the mode,  $\alpha$  being its attenuation constant and  $\beta$  its phase constant. Substituting this field dependency into the scalar wave equations, and writing them for  $TE^x$  ( $E_x = 0$ ) and  $TM^x$  ( $H_x = 0$ ) modes while making use of  $\nabla \cdot [\epsilon(x)E] = 0$  and  $\nabla \cdot H = 0$  accordingly, yields simplified and uncoupled scalar wave equations that are readily solved. The  $E_y$  component of the  $TE^x$  15 modes must satisfy the Helmholtz wave equation:

$$\frac{\partial^2}{\partial x^2} E_y^{TE} + \frac{\partial^2}{\partial y^2} E_y^{TE} + [\gamma^2 + \omega^2 \mu \epsilon(x)] E_y^{TE} = 0 \quad (4)$$

and the  $H_y$  component of the  $TM^x$  modes must satisfy the Sturm-Liouville wave equation:

$$20 \quad \epsilon(x) \frac{\partial}{\partial x} \left[ \frac{1}{\epsilon(x)} \frac{\partial}{\partial x} H_y^{TM} \right] + \frac{\partial^2}{\partial y^2} H_y^{TM} + [\gamma^2 - \omega^2 \mu \epsilon(x)] H_y^{TM} = 0 \quad (5)$$

The superposition of the  $TE^x$  and  $TM^x$  mode families then describes any mode propagating in the structure analyzed. The electric and magnetic field components resulting from this superposition are given by the following equations:

$$25 \quad E_x = \frac{-1}{j\omega\gamma} \left[ \frac{\partial}{\partial x} \left( \frac{1}{\epsilon(x)} \frac{\partial}{\partial x} H_y^{TM} \right) + \omega^2 \mu H_y^{TM} \right] \quad (6)$$

$$E_y = E_y^{TE} - \frac{i}{j\omega\gamma\epsilon(x)} \frac{\partial^2}{\partial x \partial y} H_y^{TM} \quad (7)$$

$$30 \quad E_z = \frac{1}{\gamma} \frac{\partial}{\partial y} E_y^{TE} + \frac{1}{j\omega\epsilon(x)} \frac{\partial}{\partial x} H_y^{TM} \quad (8)$$

$$35 \quad H_x = \frac{1}{j\omega\gamma} \left[ \frac{1}{\mu} \frac{\partial^2}{\partial x^2} E_y^{TE} + \omega^2 \epsilon(x) E_y^{TE} \right] \quad (9)$$

$$H_y = \frac{1}{j\omega\gamma\mu} \frac{\partial^2}{\partial x \partial y} E_y^{TE} + H_y^{TM} \quad (10)$$

$$H_z = -\frac{1}{j\omega\mu} \frac{\partial}{\partial x} E_y^{TE} + \frac{1}{\gamma} \frac{\partial}{\partial y} H_y^{TM} \quad (11)$$

In order to obtain a mode of propagation supported by a waveguiding structure, the Helmholtz and Sturm-Liouville wave equations (4) and (5), along with the field equations (6)-(11), must be solved for the propagation constant  $\gamma$  using appropriate boundary conditions applied between layers and at the horizontal and vertical limits.

5

### Poynting Vector and Power Confinement Factor

The power confinement factor is defined as the ratio of mode complex power carried through a portion of a waveguide's cross-section with respect to the mode complex power carried through the entire waveguide cross-section. Formally it is

10 expressed as:

$$cf = \frac{\left| \iint_{A_c} S_z ds \right|}{\left| \iint_{A_\infty} S_z ds \right|} \quad (12)$$

15 where  $A_c$  is usually taken as the area of the waveguide core and  $A_\infty$  implies integration over the entire waveguide cross-section (which can be all cross-sectional space for an open structure) or the entire cross-sectional computational domain.  $S_z$  refers to the z component of the Poynting vector:

20

$$S_z = \frac{1}{2} (E_x H_y^* - E_y H_x^*) \quad (13)$$

and  $H_{x,y}^*$  denotes the complex conjugate of  $H_{x,y}$ . The spatial distribution of a component of the Poynting vector is easily computed from the spatial distribution of the relevant electric and magnetic mode field components.

25

### Numerical Solution Approach

The boundary value problem governed by equations (4) to (11) is solved by applying the Method of Lines (MoL). The MoL is a well-known numerical technique and its application to various electromagnetic problems, including optical waveguiding, 30 is well-established [14]. The MoL is rigorous, accurate and flexible. It can handle a wide variety of waveguide geometries, including the structures at hand. The method is not known to generate spurious or non-physical modes. The MoL formulation used herein is based on the formulation reported in [15], but simplified for isotropic media, as prescribed by equations (4) - (11) and reported in [16]. Except for a 1-D spatial 35 discretization, the method is exact.

The main idea behind the MoL is that the differential field equations governing a waveguiding problem are discretized only as far as necessary so that generalized analytic solutions can be applied to derive a homogeneous matrix problem describing all

modes supported by the structure. This approach renders the method accurate and computationally efficient since only N-1 dimensions must be discretized to solve an N dimension problem. In the case of a two-dimensional (2-D) waveguiding structure, this means that only one spatial dimension needs to be discretized. The main features of this 5 procedure, as applied to a modal analysis problem, are described below.

- The  $x$  axis and the function  $\epsilon(x)$  are discretized using two shifted non-equidistant line systems, parallel to the  $y$  axis.
- The differential operators  $\partial/\partial x$  and  $\partial^2/\partial^2 x$  in the wave and field equations are replaced by finite difference approximations that include the lateral boundary conditions.
- The discretized wave equations are diagonalized using appropriate transformations matrices.
- The diagonalization procedure yields in the transform domain two systems of uncoupled one-dimensional (1-D) differential equations along the remaining dimension (in this case along the  $y$  axis).
- These differential equations are solved analytically and tangential field matching conditions are applied at interfaces between layers along with the top and bottom boundary conditions.
- The last field matching condition, applied near the center of the structure, yields a homogeneous matrix equation of the form  $G(\gamma)\tilde{e} = 0$  which operates on transformed tangential fields.
- The complex propagation constant  $\gamma$  of modes is then obtained by searching for values that satisfy  $\det[G(\gamma)] = 0$ .
- Once the propagation constant of a mode has been determined, the spatial distribution of all six field components of the mode are easily generated.
- A mode power confinement factor can be computed by first computing the spatial distribution of  $S_z$  which is then integrated according to Equation (12).

The open structures shown in Figures 1(a) and 17(a) are discretized along the  $x$  30 axis and the generalized analytic solution applied along the  $y$  axis. The physical symmetry of the structures is exploited to increase the accuracy of the results and to reduce the numerical effort required to generate the mode solutions. For the symmetric structure shown in Figure 1(a), this is achieved by placing either electric wall ( $E_{\text{tan}} = 0$ ) or magnetic wall ( $H_{\text{tan}} = 0$ ) boundary conditions along the  $x$  and  $y$  axes. For the 35 asymmetric structure shown in Figure 17(a), this is achieved by placing electric wall or magnetic wall boundary conditions along the  $y$  axis only. The remaining horizontal boundary conditions are placed at infinity and the remaining lateral boundary condition is either placed far enough from the guide to have a negligible effect on the mode

calculation, or a lateral absorbing boundary condition is used to simulate infinite space, depending on the level of confinement observed in the resulting mode. The use of numerical methods to solve differential equations inevitably raises questions regarding the convergence of computed results and their accuracy. The propagation constant of 5 a mode computed using the method of lines converges in a monotonic or smooth manner with a reduction in the discretization interval (which increases the number of lines in the calculation and thus the numerical effort). This suggests that extrapolation can be used to generate a more accurate value for the propagation constant, and this value can then be used to compute the error in values obtained using the coarser discretizations [17]. 10 This *anticipated error* does not correspond to the actual error in the propagation constant as the latter could only be known if the analytic or exact value were available. The anticipated error, however, still provides a useful measure of accuracy since it must tend toward zero as more accurate results are generated.

The convergence of the computed propagation constant of the modes supported 15 by the structures of interest has been monitored during the entire study. The anticipated error in the results presented herein is estimated as 1% on average and 6% in the worst case. These error values are based on extrapolated propagation constants computed using Richardson's extrapolation formula [18].

20 III. Mode Characteristics and Evolution With Film Thickness: Symmetric Structures  
A. Review of Mode Solutions for Metal Film Slab Waveguides

The review begins with the reproduction of results for an infinitely wide symmetric metal film waveguide, as shown in Figure 1(a) with  $w = \infty$ , taken from the standard work on such structures [6]. In order to remain consistent with their results, 25 the optical free-space wavelength of excitation is set to  $\lambda_0 = 0.633\mu\text{m}$  and their value for the relative permittivity of the silver film at this wavelength is used:  $\epsilon_{r,2} = -19 - j0.53$ . The relative permittivity of the top and bottom dielectric regions is set to  $\epsilon_{r,1} = 4$ .

An infinitely wide structure supports only two purely bound TM ( $E_x = H_y = H_z = 0$ ) modes having transverse field components  $E_y$  and  $H_x$  that exhibit asymmetry or symmetry with respect to the  $x$  axis. These modes are created from the coupling of individual plasmon-polariton modes supported by the top and bottom interfaces and they exhibit dispersion with film thickness. The widely accepted nomenclature for identifying them consists in using the letters *a* or *s* for asymmetric or symmetric transverse field 30 distributions, respectively, followed by a subscript *b* or *l* for bound or leaky modes, respectively. The propagation constants of the  $a_b$  and  $s_b$  modes have been computed as 35 a function of film thickness and the normalized phase and attenuation constants are plotted in Figures 2(a) and 2(b), respectively.

From Figures 2(a) and 2(b), it is observed that the  $a_b$  and  $s_b$  modes become degenerate with increasing film thickness. As the separation between the top and bottom interfaces increases, the  $a_b$  and  $s_b$  modes begin to split into a pair of uncoupled plasmon-polariton modes localized at the metal-dielectric interfaces. The propagation constants of the  $a_b$  and  $s_b$  modes thus tend towards that of a plasmon-polariton mode supported by the interface between semi-infinite metallic and dielectric regions, which is given via the following equations [6]:

10

$$\beta/\beta_0 = -Re \left\{ \sqrt{\frac{\epsilon_{r,1}\epsilon_{r,2}}{\epsilon_{r,1} + \epsilon_{r,2}}} \right\} \quad (14)$$

$$\alpha/\beta_0 = -Im \left\{ \sqrt{\frac{\epsilon_{r,1}\epsilon_{r,2}}{\epsilon_{r,1} + \epsilon_{r,2}}} \right\} \quad (15)$$

15 where  $\beta_0 = \omega/c_0$  with  $c_0$  being the velocity of light in free space, and  $\epsilon_{r,1}$  and  $\epsilon_{r,2}$  are the complex relative permittivities of the materials used. Using the above equations, values of  $\beta/\beta_0 = 2.250646$  and  $\alpha/\beta_0 = 0.836247 \times 10^{-2}$  are obtained for  $\epsilon_{r,1} = 4$  and  $\epsilon_{r,2} = -19 - j 0.53$ .

As the thickness of the film decreases, the phase and attenuation constants of the  $a_b$  mode increase, becoming very large for very thin films. This is due to the fact that the fields of this mode penetrate progressively deeper into the metal as its thickness is reduced. In the case of the  $s_b$  mode, a decreasing film thickness causes the opposite effect, that is, the fields penetrate progressively more into the top and bottom dielectric regions and less into the metal. The propagation constant of this mode thus tends asymptotically towards that of a TEM (Transverse ElectroMagnetic) wave propagating in an infinite medium having the same permittivity as the top and bottom dielectric regions. In this case, the attenuation constant decreases asymptotically towards zero since losses were neglected in these regions. The  $a_b$  and  $s_b$  modes do not have a cutoff thickness.

20 The fields in an infinitely wide structure do not exhibit any spatial variation along  $x$ . Due to the nature of the MoL, and to the fact that the generalized analytical solution is applied along the  $y$  dimension, our results do not contain discretization errors and thus are in perfect agreement with those reported in [6].

30 **35 B. Modes Supported by a Metal Film of Width  $w = 1\mu m$**

Next, the analysis of the structure shown in Figure 1(a) for the case  $w = 1\mu m$  will be explained. The material parameters and free-space wavelength that were used in the previous case ( $w = \infty$ ) were also used here. The MoL was applied and the

discretization adjusted until convergence of the propagation constant was observed. The physical quarter-symmetry of the structure was exploited by placing vertical and horizontal electric or magnetic walls along the  $y$  and  $x$  axes, respectively, which leads to four possible wall combinations as listed in Table 1. The first two purely bound (non-leaky) modes for each wall combination were found and their dispersion with metal thickness computed. The results for these eight modes are shown in Figures 2(a) and 2(b).

Table 1: Vertical-Horizontal wall combinations used along the axes of symmetry and proposed mode nomenclature: ew - electric wall, mw - magnetic wall.

15

V-H Walls	Mode
ew-ew	$as_b^m$
mw-ew	$ss_b^m$
mw-mw	$sa_b^m$
ew-mw	$aa_b^m$

Unlike its slab counterpart, pure TM modes are not supported by a metal film of finite width: all six field components are present in all modes. For a symmetric structure having an aspect ratio  $w/t > 1$ , the  $E_y$  field component dominates. The  $E_x$  field component increases in magnitude with increasing film thickness and if  $w/t < 1$ , then  $E_x$  dominates. It is proposed to identify the modes supported by a metal film of finite width, by extending the nomenclature used for metal film slab waveguides. First a pair of letters being  $a$  or  $s$  identify whether the main transverse electric field component is asymmetric or symmetric with respect to the  $y$  and  $x$  axes, respectively (in most practical structures  $w/t \gg 1$  and  $E_y$  is the main transverse electric field component). A superscript is then used to track the number of extrema observed in the spatial distribution of this field component along the largest dimension (usually along the  $x$  axis) between the corners. A second superscript  $n$  could be added to track the extrema along the other dimension (the  $y$  axis) if modes exhibiting them are found. Finally, a subscript  $b$  or  $l$  is used to identify whether the mode is bound or leaky. Leaky modes are known to exist in metal film slab structures and though a search for them has not been made at this time, their existence is anticipated. Table 1 relates the proposed mode nomenclature to the corresponding vertical and horizontal wall combinations used along the axes of symmetry.

35 The  $ss_b^0$ ,  $sa_b^0$ ,  $as_b^0$  and  $aa_b^0$  modes are the first modes generated (one for each of the four possible quarter-symmetries listed in Table 1, and having the largest phase

constant) and thus may be considered as the fundamental modes supported by the structure. Figures 3 to 6 show the field distributions of these modes over the cross-section of the waveguide for a metal film of thickness  $t = 100\text{nm}$ . As is observed from these figures, the main transverse electric field component is the  $E_y$  component and the symmetries in the spatial distribution of this component are reflected in the mode nomenclature. The outline of the metal is clearly seen in the distribution of the  $E_y$  component on all of these plots. As is observed from the figures, very little field tunnels through the metal to couple parallel edges for this case of film thickness and width (very little coupling through the metal between the top and bottom edges and between the left and right edges), though coupling does occur along all edges between adjacent corners (mostly along the left and right ones), and also between perpendicular edges through the corner.

Figures 2(a) and 2(b) suggests that the dispersion curves for these first four modes converge with increasing film thickness toward the propagation constant of a plasmon-polariton mode supported by an isolated corner (though pairs of corners in this case remain weakly coupled along the top and bottom edges due to the finite width of the film, even if its thickness goes to infinity). If both the film thickness and width were to increase further, the four fundamental modes would approach degeneracy with their propagation constant tending towards that of a plasmon-polariton mode supported by an isolated corner, and their mode fields becoming more localized near the corners of the structure with maxima occurring at all four corners and fields decaying in an exponential-like manner in all directions away from the corners. This is further supported by considering the evolution of the field distributions given in Figures 3 to 6 as both the thickness and width increase.

As the thickness of the film decreases, coupling between the top and bottom edges increases and the four modes split into a pair as the upper branch (modes  $sa_b^0$  and  $aa_b^0$  which have a dominant  $E_y$  field component exhibiting asymmetry with respect to the  $x$  axis) and a pair as the lower branch (modes  $ss_b^0$  and  $as_b^0$  which have a dominant  $E_y$  field component exhibiting symmetry with respect to the  $x$  axis), as shown in Figures 2(a) and 2(b). The pair on the upper branch remain approximately degenerate for all film thicknesses, though decreasing the film width would eventually break this degeneracy. The upper branch modes do not change in character as the film thickness decreases. Their field distributions remain essentially unchanged from those shown in Figures 4 and 6 with the exception that confinement to the metal region is increased thus causing an increase in their attenuation constant. This field behaviour is consistent with that of the  $a_b$  mode supported by a metal film slab waveguide.

The modes on the lower branch begin to split at a film thickness of about  $80\text{nm}$ , as shown in Figures 2(a) and 2(b). As the film thickness decreases further

the  $ss_b^0$  mode follows closely the phase and attenuation curves of the  $s_b$  mode supported by the metal film slab waveguide. In addition to exhibiting dispersion, the lower branch modes change in character with decreasing thickness, their fields evolving from being concentrated near the corners, to having Gaussian-like distributions along the waveguide width. The  $E_y$  field component of the  $ss_b^0$  mode develops an extremum near the center of the top and bottom interfaces, while that of the  $as_b^0$  mode develops two extrema, one on either side of the center. Since these modes change in character, they should be identified when the film is fairly thick.

Figures 7(a) to 7(f) show the evolution of the  $ss_b^0$  mode fields with film thickness via contour plots of  $Re\{S_z\}$ .  $S_z$  is computed from the  $ss_b^0$  mode fields using Equation 13 and corresponds to the complex power density carried by the mode. The power confinement factor  $cf$  is also given in the figure for all cases, and is computed via equation (12) with the area of the waveguide core  $A_c$  taken as the area of the metal region. Figures 7(a) to 7(f) clearly show how the mode fields evolve from being confined to the corners of thick films to being distributed in a Gaussian-like manner laterally along the top and bottom edges, as the field coupling between these edges increases due to a reduction in film thickness. The confinement factor becomes smaller as the film thickness decreases, ranging from 14% confinement to 1.6% as the thickness goes from 80nm to 20nm. This implies that fields become less confined to the metal, spreading out not only along the vertical dimension but along the horizontal one as well, as is observed by comparing Figures 7(a) and 7(b). This reduction in confinement to the lossy metal region explains the reduction in the attenuation constant of the mode with decreasing film thickness, as shown in Figure 2(b). An examination of all field components related to the  $ss_b^0$  mode reveals that the magnitudes of the weak transverse ( $E_x, H_y$ ) and longitudinal ( $E_z, H_z$ ) components decrease with decreasing film thickness, implying that the mode is evolving towards a TEM mode comprised of the  $E_y$  and  $H_x$  field components. Indeed, the normalized propagation constant of the  $ss_b^0$  mode tends asymptotically towards the value of the normalized propagation constant of a TEM wave propagating in the background material ( $\epsilon_{r,1} = 4$  with no losses in this case), further supporting this fact. This field behaviour is also consistent with that of the  $s_b$  mode supported by a metal film slab waveguide.

Figure 8 shows the profile of  $Re\{S_z\}$  of the  $ss_b^0$  mode over the cross-section of the guide for the case  $t = 20nm$ , providing a different perspective of the same information plotted as contours in Figure 7(f). Figure 8 shows that  $Re\{S_z\}$  is negative in the metal film, implying that the mode real power is flowing in the direction opposite to the direction of mode propagation (or to the direction of phase velocity) in this region. It is clear however that the overall or net mode real power is flowing along the direction of propagation. It is possible that the net mode real power can be made to flow in the

direction opposite to that of phase velocity (as in metal film slab waveguides [10]) for values of  $\epsilon_{r,1}$  in the neighbourhood or greater than  $|Re\{\epsilon_{r,2}\}|$ .

Unlike the metal film slab waveguide, a metal film of finite width can support a number of higher order modes. The dispersion curves of the first four higher order modes (each generated from one of the symmetries listed in Table 1) are shown in Figures 2(a) and 2(b), and the spatial distribution of their main transverse electric field component is shown in Figure 9 for a film of thickness  $t = 100nm$ . As is observed from Figures 9(a) to 9(d), the symmetries and number of extrema in the distributions of  $Re\{E_y\}$  are reflected in the mode nomenclature. It should be noted that the nature of the nomenclature is such that all higher order modes  $sa_b^m$  and  $ss_b^m$  have an odd  $m$  while all higher order modes  $aa_b^m$  and  $as_b^m$  have an even  $m$ . Comparing Figures 9(a) to 9(d) with Figures 3(c), 4(c), 5(c) and 6(c), respectively, (ie: comparing the  $E_y$  component of the  $ss_b^1$  mode shown in Figure 9(a) with the  $E_y$  component of the  $ss_b^0$  mode shown in Figure 3(c), etc...) reveals that the fields of a higher order mode are comprised of the fields of the corresponding  $m = 0$  mode with additional spatial oscillations or variations along the top and bottom edges of the structure due to the latter's limited width. Making this comparison for all of the field components of the higher order modes found reveals this fact to be true, except for the  $H_y$  field component which remains in all cases essentially identical to that of the corresponding  $m = 0$  mode; ie: the  $H_y$  field component never exhibits oscillations along the width of the structure.

The evolution of the  $sa_b^1$  and  $aa_b^2$  modes with film thickness is similar to the evolution of the  $sa_b^0$  and  $aa_b^0$  modes (and the  $a_b$  mode supported by the metal film slab waveguide), in that their mode fields become more tightly confined to the metal as the thickness of the latter decreases, thereby causing an increase in the attenuation of the modes, as shown in Figure 2(b). Furthermore, the  $sa_b^1$  and  $aa_b^2$  modes do not change in character with film thickness, their field distributions remaining essentially unchanged in appearance from those computed at a thickness of  $100nm$ .

The  $ss_b^1$  and  $as_b^2$  modes evolve with thickness in a manner similar to the corresponding  $m = 0$  modes (and the  $s_b$  mode of the metal film slab waveguide) in the sense that their fields become less confined to the metal region as the thickness of the latter decreases, thereby reducing the attenuation of the modes as shown in Figure 2(b). As the thickness of the film decreases, the  $ss_b^1$  and  $as_b^2$  modes change in character in a manner similar to the corresponding  $m = 0$  modes, their field components evolving extra variations along the top and bottom edges.

As the thickness of the film increases, the propagation constants of the  $sa_b^1$  and  $ss_b^1$  modes converge to a single complex value as shown in Figures 2(a) and 2(b). This is the propagation constant of uncoupled higher order modes supported by the top and bottom edges of the film. A similar observation holds for

the  $aa_b^2$  and  $as_b^2$  modes. The nature of these 'edge modes' is clear by considering the evolution with increasing film thickness of the distributions shown in Figures 9(a) to 9(d). As the thickness of the film tends to infinity, the top edge becomes uncoupled from the bottom edge, forcing the  $ss_b^1$  mode to become degenerate with the  $sa_b^1$  mode 5 since both have an  $E_y$  field component that is symmetric with respect to the  $y$  axis and one extremum in its distribution along the top or bottom edge. A similar reasoning explains why the  $as_b^2$  mode must become degenerate with the  $ss_b^2$  mode. In general, it is expected that the higher order  $sa_b^m$  and  $ss_b^m$  mode families will form degenerate pairs for a given  $m$ , as will the higher order  $as_b^m$  and  $aa_b^m$  mode families, with 10 increasing film thickness.

The  $aa_b^m$  and  $sa_b^m$  mode families do not have mode cutoff thicknesses. This is due to the fact that their confinement to the metal film increases with decreasing film thickness; thus the modes remain guided as  $t \rightarrow 0$ . The  $as_b^m$  and  $ss_b^m$  mode families have cutoff thicknesses for all modes except the  $ss_b^0$  mode, which remains guided as  $t$  15  $\rightarrow 0$ , since it evolves into the TEM mode supported by the background. The other modes of these families, including the  $as_b^0$  mode cannot propagate as  $t \rightarrow 0$  because their mode fields do not evolve into a TEM mode. Rather, the modes maintain extrema in their field distributions and such variations cannot be enforced by an infinite homogeneous medium.

In general, the purely bound modes supported by a metal film of finite width appear to be formed from a coupling of modes supported by each metal-dielectric interface defining the structure. In a metal film of finite width, straight interfaces of finite length (top, bottom, left and right edges) and corner interfaces are present. Since a straight metal-dielectric interface of infinite length can support a bound plasmon-polariton mode then so should an isolated corner interface and a straight interface of finite length bounded by corners (say the edge defined by a metal of finite width having an infinite thickness). A preliminary analysis of an isolated corner has revealed that a plasmon-polariton mode is indeed supported and that the phase and attenuation constants of this mode are greater than those of the mode guided by the corresponding infinite 25 straight interface, as given by Equations (14) and (15). This is due to the fact that fields penetrate more deeply into the metal near the corner, to couple neighbouring perpendicular edges. All six field components are present in such a mode, having their maximum value at the corner and decreasing in an exponential-like manner in all directions away from the corner. A straight interface of finite length bounded by corners 30 should support a discrete spectrum of plasmon-polariton modes with the defining feature in the mode fields being the number of extrema in their spatial distribution along the edge. A mode supported by a metal film of finite width may therefore be seen as being 35 comprised of coupled 'corner modes' and 'finite length edge modes'.

The  $ss_b^0$  mode could be used for optical signal transmission over short distances. Its losses decrease with decreasing film thickness in a manner similar to the  $s_b$  mode supported by the metal film slab waveguide. In a symmetric waveguide structure such as the one studied here, the  $ss_b^0$  mode does not have a cut-off thickness so losses could 5 be made small enough to render it long-ranging, though a trade-off against confinement is necessary. In addition, when the metal is thin, the  $E_y$  field component of the mode has a maximum near the center of the metal-dielectric interfaces, with a symmetric profile similar to that shown in Figure 8. This suggests that the mode should be excitable using a simple end-fire technique similar to the one employed to excite surface 10 plasmon-polariton modes [19,6]; this technique is based on maximizing the overlap between the incident field and that of the mode to be excited.

In reference [22], the present inventor *et al.* disclosed that plasmon-polariton waves supported by thin metal films of finite width have recently been observed experimentally at optical communications wavelengths using this method of excitation.

15

#### IV. Mode Dispersion With Film Width: Symmetric Structures

Since the modes supported by a metal film waveguide exhibit dispersion with film thickness, it is expected that they also exhibit dispersion with film width.

##### 20 A. Modes Supported by a Metal Film of Width $w = 0.5\mu m$

The analysis of a metal film waveguide of width  $w = 0.5\mu m$  will now be discussed, using the material parameters and free-space wavelength that were used in the previous section. A film width of  $0.5\mu m$  was selected in order to determine the impact of a narrowing film on the modes supported and to demonstrate that the structure can 25 still function as a waveguide though the free-space optical wavelength is greater than both the width and thickness of the film.

As in the previous section, the first eight modes supported by the structure (two for each symmetry listed in Table 1) were sought, but in this case only six modes were found. The dispersion curves with thickness of the modes found are plotted in Figures 30 10(a) and 10(b). The observations made in the previous section regarding the general behaviour of the modes hold for other film widths, including this one.

The  $aa_b^2$  and  $as_b^2$  modes, which were the highest order modes found for a film of width  $w = 1\mu m$ , were not found in this case suggesting that the higher order modes ( $m > 0$ ) in general have a cut-off width. Comparing Figure 10(a) with Figure 2(a), it 35 is apparent that decreasing the film width causes a decrease in the phase constant of the  $ss_b^1$  and  $sa_b^1$  modes, further supporting the existence of a cut-off width for these modes.

Comparing Figures 10(a) and 10(b) with Figures 2(a) and 2(b), it is noted that the modes which do exhibit cutoff thicknesses (the  $ss_b^m$  modes with  $m > 0$  and the  $as_b^m$  modes with  $m \geq 0$ ), exhibit them at a larger thickness for a narrower film width. This makes it possible to design a waveguide supporting only one long-ranging mode 5 (the  $ss_b^0$  mode) by carefully selecting the film width and thickness.

### B. Dispersion of the $ss_b^0$ Mode With Film Width

The dispersion with thickness of the  $ss_b^0$  mode is shown in Figures 11(a) and 11(b) for numerous film widths in the range  $0.25 \leq w \leq 1\mu m$ , illustrating the amount 10 of dispersion in the mode properties that can be expected due to a varying film width. In all cases, the  $ss_b^0$  mode evolves with decreasing film thickness into the TEM wave supported by the background, but this evolution occurs more rapidly for a narrower width. For a film of thickness  $t = 20nm$ , for example, from Figure 11(a), the normalized phase constant of the mode supported by a film of width  $w = 1\mu m$  is about 15 2.05, while that of the mode supported by a film of width  $w = 0.25\mu m$  is already about 2. This fact is also supported by the results plotted in Figure 11(b) since the attenuation constant of the mode at a thickness of  $t = 20nm$  is closer to zero (the attenuation constant of the background) for narrow film widths compared to wider ones. Indeed, at a thickness of  $10nm$ , the attenuation of the mode for a width of  $w = 0.25\mu m$  is more 20 than an order of magnitude less than its attenuation at a width of  $w = 1\mu m$  (and more than an order of magnitude less than that of the  $s_b$  mode supported by a metal film slab waveguide), indicating that this mode can be made even more long-ranging by reducing both the film thickness and its width.

The dispersion of the mode with increasing film thickness also changes as a 25 function of film width, as seen from Figure 11(a). This is due to the fact that the amount of coupling between corners along the top and bottom edges increases as the film narrows, implying that the mode does not evolve with increasing thickness towards a plasmon-polariton mode supported by an isolated corner, but rather towards a plasmon-polariton mode supported by the pair of corners coupled via these edges.

30 Figures 12(a) to 12(d) show contour plots of  $Re\{S_z\}$  related to the  $ss_b^0$  mode supported by films of thickness  $t = 20nm$  and various widths. The power confinement factor is also given for all cases, with the area of the waveguide core  $A_c$  taken as the area of the metal region. Figures 12(a) to 12(d) clearly illustrate how the fields become less confined to the lossy metal as its width decreases, explaining the reduction in 35 attenuation shown in Figure 11(b) at this thickness. In addition, the confinement factor ranges from 1.64% to 0.707% for the widths considered, further corroborating this fact. The fields are also seen to spread out farther, not only along the horizontal dimension but along the vertical one as well, as the film narrows. This indicates that the mode

supported by a narrow film is farther along in its evolution into the TEM mode supported by the background, compared to a wider film of the same thickness. It is also clear from Figures 12(a) to 12(d) that the trade-off between mode confinement and attenuation must be made by considering not only the film thickness but its width as  
5 well.

#### V. Effects Caused by Varying the Background Permittivity: Symmetric Structures

In this section, the changes in the propagation characteristics of the  $ss_b^0$  mode due to variations in the background permittivity of the waveguide. Only the  $ss_b^0$  mode  
10 is considered since the main effects are in general applicable to all modes. In order to isolate the effects caused by varying the background permittivity, the width of the metal film was fixed to  $w = 0.5\mu m$  and its permittivity as well as the optical free-space wavelength of analysis were set to the values used in the previous sections. The relative permittivity of the background  $\epsilon_{r,1}$  was taken as the variable parameter.  
15 The dispersion with thickness of the  $ss_b^0$  mode is shown in Figure 13 for some background permittivities in the range  $1 \leq \epsilon_{r,1} \leq 4$ . Figures 14(a) to 14(d) compare contour plots of  $Re\{S_z\}$  related to this mode for a film of thickness  $t = 20nm$  and for the same set of background permittivities used to generate the curves plotted in Figure 13. From Figures 14(a) to 14(d), it is observed that reducing the value of the background  
20 permittivity causes a reduction in field confinement to the metal. This reduction in field confinement within the lossy metal in turn causes a reduction in the attenuation of the mode that can be quite significant, Figure 13 showing a reduction of almost four orders of magnitude at a film thickness of  $t = 20nm$ , as the background relative permittivity ranges from  $\epsilon_{r,1} = 4$  to 1. It is also noted that the mode exhibits less dispersion with  
25 thickness as the background relative permittivity is reduced, since the normalized phase constant curves shown in Figure 13 flatten out with a reduction in the value of this parameter.

From Figures 14(a) to 14(d), it is seen that the mode power is confined to within approximately one free-space wavelength in all directions away from the film in all cases  
30 except that shown in Figure 14(d), where fields are significant up to about two free-space wavelengths. In Figure 14(c), the background permittivity is roughly that of glass and from Figure 13 the corresponding normalized attenuation constant of the mode is about  $\alpha/\beta_0 = 6.0 \times 10^{-5}$ . The associated mode power attenuation in  $dB/mm$ , computed using the following formula:

35

$$Att = \alpha \times \frac{20}{1000} \log_{10}(e) \quad (16)$$

is about 5 dB/mm. This value of attenuation is low enough and field confinement is high enough as shown in Figure 14(c), to render this particular structure practical at this free-space wavelength for applications requiring short propagation lengths.

The changes in mode properties caused by varying the background permittivity

5 as discussed above are consistent with the changes observed for the modes supported by a metal film slab waveguide and the observations are in general applicable to the other modes supported by a metal film of finite width. In the case of the higher order modes ( $m > 0$ ) and those exhibiting a cutoff thickness (the  $as_b^m$  modes for all  $m$  and the  $ss_b^m$  modes for  $m > 0$ ) additional changes in the mode properties occur. In  
10 particular, as the background permittivity is reduced, the cut-off widths of the higher order modes increase as do all relevant cut-off thicknesses.

#### VI. Frequency Dependency of the $ss_b^0$ Mode Solutions: Symmetric Structures

In order to isolate the frequency dependency of the  $ss_b^0$  mode solutions, the  
15 geometry of the metal film was held constant and the background relative permittivity was set to  $\epsilon_{r,1} = 4$ . The relative permittivity of the metal film  $\epsilon_{r,2}$  was assumed to vary with the frequency of excitation according to Equation (1). In order to remain consistent with [6], the values  $\omega_p = 1.29 \times 10^{16} \text{ rad/s}$  and  $1/\nu = \tau = 1.25 \times 10^{-14} \text{ s}$  were adopted, though the latter do not generate exactly  $\epsilon_{r,2} = -19 - j0.53$  at  $\lambda_0 = 0.633 \mu\text{m}$ , which is  
20 the value used in the previous sections. This is due to the fact that values of  $\omega_p$  and  $\tau$  are often deduced by fitting Equation (1) to measurements. The values used, however, are in good agreement with recent measurements made for silver [3] and are expected to generate frequency dependent results that are realistic and experimentally verifiable.

The dispersion characteristics of the  $ss_b^0$  mode supported by films of width  $w =$   
25  $0.5 \mu\text{m}$  and  $w = 1 \mu\text{m}$ , and thicknesses in the range  $10 \leq t \leq 50 \text{ nm}$  are shown in Figures 15(a) and 15(b) for frequencies covering the free-space wavelength range  $0.5 \leq \lambda_0 \leq 2 \mu\text{m}$ . Curves for the  $s_b$  mode supported by metal film slab waveguides ( $w = \infty$ ) of the same thicknesses are also shown for comparison.

The results given in Figure 15(a) show that, in all cases, the normalized phase  
30 constant of the modes tends asymptotically towards that of the TEM wave supported by the background as the wavelength increases, and that the convergence to this value is steeper as the width of the film decreases (for a given thickness). The curves remain essentially unchanged in character as the thickness changes, but they shift upwards toward the top left of the graph with increasing thickness, as shown. Convergence to  
35 the asymptote value with increasing wavelength suggests that the  $ss_b^0$  mode evolves into the TEM mode supported by the background. It is noteworthy that the  $ss_b^0$  mode can exhibit very little dispersion over a wide bandwidth, depending on the thickness and

width of the film, though flat dispersion is also associated with low field confinement to the metal film.

The results plotted in Figure 15(b) show in all cases a decreasing attenuation with increasing wavelength and the curves show a sharper drop for a narrow film ( $w = 0.5\mu m$ ) compared to a wide one ( $w = \infty$ ). The attenuation curves look essentially the same for all of the film thicknesses considered, though the range of attenuation values shifts downwards on the graph with decreasing film thickness.

Figures 16(a) to 16(f) give contour plots of  $Re\{S_z\}$  related to the  $ss_b^0$  mode for films of thickness  $t = 20nm$  and widths  $w = 0.5\mu m$  and  $w = 1\mu m$ , for three free-space wavelengths of operation:  $\lambda_0 = 0.6, 0.8$  and  $1.2\mu m$ . Comparing the contours shown in Figures 16(a) to 16(f), explains in part the frequency dependent behaviour plotted in Figures 15(a) and 15(b). Figures 16(a) to 16(f) show that the mode power contours spread out farther from the film as the wavelength increases, which means that the mode confinement to the metal region decreases, explaining in part the decrease in losses and the evolution of the mode towards the TEM mode of the background, as shown in Figures 15(a) and 15(b). This behaviour is more pronounced for the waveguide of width  $w = 0.5\mu m$  compared to the wider one of width  $w = 1.0\mu m$ .

There are two mechanisms causing changes in the  $ss_b^0$  mode as the frequency of operation varies. The first is geometrical dispersion, which changes the optical or apparent size of the film, and the second is material dispersion, which is modeled for the metal region using Equation (1). If no material dispersion is present, then the geometrical dispersion renders the film optically smaller as the free-space wavelength is increased (an effect similar to reducing  $t$  and  $w$ ) so, in the case of the  $ss_b^0$  mode, confinement to the film is reduced and the mode spreads out in all directions away from the latter. Now based on Equation (1), it is clear that the magnitude of the real part of the film's permittivity  $|Re\{\epsilon_{r,2}\}|$  varies approximately in a  $1/\omega^2$  or  $\lambda_0^2$  fashion while the magnitude of its imaginary part  $|Im\{\epsilon_{r,2}\}|$  varies approximately in a  $1/\omega^3$  or  $\lambda_0^3$  fashion. However, an increase in  $|Re\{\epsilon_{r,2}\}|$  reduces the penetration depth of the mode fields into the metal region and, combined with the geometrical dispersion, causes a net decrease in mode attenuation with increasing wavelength, even though the losses in the film increase in a  $\lambda_0^3$  fashion.

Figure 15(b) shows that mode power attenuation values in the range 10 to 0.1 dB/cm are possible near communications wavelengths ( $\lambda_0 \sim 1.5\mu m$ ) using structures of reasonable dimensions:  $w \sim 1.0\mu m$  and  $t \sim 15nm$ . Such values of attenuation are low enough to consider the  $ss_b^0$  mode as being long-ranging, suggesting that these waveguides are practical for applications requiring propagation over short distances. As shown in the previous section, even lower attenuation values are possible if the background permittivity is lowered. From Figures 16(e) and 16(f), (case  $\lambda_0 = 1.2\mu m$ ,

which is near communications wavelengths), it is apparent that the mode power confinement is within one free-space wavelength of the film, which should be tight enough to keep the mode bound to the structure if a reasonable quality metal film of the right geometry can be constructed.

5

## VII. Mode Characteristics and Evolution With Film Thickness: Small Asymmetry

### A. Mode Solutions for a Metal Film Slab Waveguide

Effects on waveguiding characteristics of using an asymmetric waveguide structure will now be discussed, beginning with the reproduction of results for an 10 infinitely wide asymmetric metal film waveguide (similar to that shown in Figure 17(a) but with  $w = \infty$ ), taken from the standard work on such structures [6]. In order to remain consistent with their results, the optical free-space wavelength of excitation is set to  $\lambda_0 = 0.633 \mu\text{m}$  and the value they used for the relative permittivity of the silver film at this wavelength is used here:  $\epsilon_{r,2} = -19 - j0.53$ . The relative permittivities of the 15 bottom and top dielectric regions are set to  $\epsilon_{r,1} = 4$  ( $n_1 = 2$ ) and  $\epsilon_{r,3} = 3.61$  ( $n_3 = 1.9$ ); these values create a structure having a small asymmetry with respect to the horizontal dimension.

The dispersion curves of the  $s_b$  and  $a_b$  modes supported by the infinitely wide structure were computed using the MoL and the results are shown in Figures 18(a) and 20 18(b). From these figures, it is seen that the propagation constant of the  $a_b$  mode tends towards that of the plasmon-polariton mode supported by the bottom interface, given by Equations (14) and (15), as the thickness of the film increases. It is also noted that this mode does not exhibit a cutoff thickness, while it is clear that the  $s_b$  mode has one near 25  $t = 18 \text{ nm}$ . The propagation constant of the  $s_b$  mode is seen to tend towards the value of a plasmon-polariton mode supported by the top interface as the thickness increases. These results are in perfect agreement with those reported in [6].

### B. Modes Supported by a Metal Film of Width $w = 1 \mu\text{m}$

The study proceeds with the analysis of the structure shown in Figure 17(a) for 30 the case  $w = 1 \mu\text{m}$ . The material parameters and free-space wavelength that were used in the previous case  $w = \infty$  were also used here. The dispersion curves for the first seven modes were computed using the MoL and the results are shown in Figures 18(a) and 18(b).

In this asymmetric structure, true field symmetry exists only with respect to the 35  $y$  axis. With respect to the horizontal dimension, the modes have a symmetric-like or asymmetric-like field distribution with field localization along either the bottom or top metal-dielectric interface. The modes that have a symmetric-like distribution with respect to the horizontal dimension are localized along the metal-dielectric interface with the

lowest dielectric constant, while modes that have an asymmetric-like distribution with respect to this axis are localized along the metal-dielectric interface with the highest dielectric constant. This behaviour is consistent with that observed for asymmetric metal slab waveguides.

5       The mode nomenclature adopted for symmetric structures can be used without ambiguity to describe the modes supported by asymmetric structures as long as the modes are identified when the metal film is fairly thick, before significant coupling begins to occur through the metal film, and while the origin of the mode can be identified unambiguously. As the metal film thickness decreases, the modes (and their  
10 fields) can evolve and change considerably more in an asymmetric structure compared to a symmetric one. The number of extrema in the main transverse electric field component of the mode is counted along the lateral dimension at the interface where the fields are localized. This number is then used in the mode nomenclature.

It was observed in Section III that the modes supported by symmetric structures  
15 are in fact supermodes created from a coupling of "edge" and "corner" modes supported by each metal-dielectric interface defining the structure. As the thickness and width of the metal decrease, the coupling between these interface modes intensifies leading to dispersion and possibly evolution of the supermode. In asymmetric structures, the bound modes are also supermodes created in a similar manner, except that dissimilar interface  
20 modes may couple to each other to create the supermode. For instance, a mode having one field extremum along the top interface (along the top edge bounded by the corners) may couple with a mode having three extrema along the bottom interface. The main selection criterion determining which interface modes will couple to create the supermode is similarity in the value of their propagation constants. For all modes supported by an  
25 asymmetric structure, an apparent symmetry or asymmetry with respect to the horizontal dimension can still be observed in the corner modes.

The  $sa_b^0$ ,  $aa_b^0$ ,  $ss_b^0$  and  $as_b^0$  modes are the fundamental modes supported by the structure. The  $sa_b^0$  and  $aa_b^0$  modes are comprised of coupled corner modes, resembling the corresponding modes in a symmetric structure, except that the fields are localized  
30 near the substrate. These two modes do not change in character as the thickness of the film decreases. A narrowing of the metal film would eventually break the degeneracy observed in Figures 18(a) and 18(b).

For a sufficiently large thickness (about 100 nm for the present structure), the  $ss_b^0$  and  $as_b^0$  modes are comprised of coupled corner modes much like the  
35 corresponding modes in a symmetric structure except that the fields are localized near the superstrate. As the thickness of the metal film decreases, both of these modes begin to evolve, changing completely in character for very thin films. Figures 19(a) to 19(d) show the evolution of the  $E_y$  field component related to the  $ss_b^0$  mode as the thickness

of the film ranges from 100 nm (Figure 19(a)) to 40 nm (Figure 19(d)). It is clearly seen that the mode evolves from a symmetric-like mode having fields localized near the superstrate to an asymmetric-like mode having fields localized along the substrate-metal interface. A similar evolution is observed for the  $as_b^0$  mode. This change in character  
5 is also apparent in their dispersion curves: they follow the general behaviour of a symmetric-like mode for large thicknesses but then slowly change to follow the behaviour of an asymmetric-like mode as the thickness decreases. Since the substrate dielectric constant is larger than the superstrate dielectric constant, the mode is "pulled"  
10 from a symmetric-like mode to an asymmetric-like mode (having field localization at the substrate-metal interface) as the metal film becomes thinner.

Figures 20(a) to 20(d) show the  $E_y$  field component related to the  $ss_b^1$  and  $sa_b^1$  modes for two film thicknesses. From these Figures it is noted that the top and bottom edge modes comprising a supermode are different from each other. In Figure 20(a), for instance, it is seen that the bottom edge mode has three extrema and  
15 is of higher order than the top edge mode which has one extremum. A similar observation holds for Figure 20(c), where it can be seen that the bottom edge mode has one extremum while the top one has none. In this structure, the substrate has a higher dielectric constant than the superstrate so the phase constant of a particular substrate-metal interface mode will be higher than the phase constant of the same mode at the  
20 metal-superstrate interface. Since a supermode is created from a coupling of edge modes having similar propagation constants, it should be expected that, in an asymmetric structure, different edge modes may couple to create a supermode. In general, higher-order modes have smaller values of phase constant compared to lower-order modes, so in structures having  $\epsilon_3 < \epsilon_1$ , all supermodes are comprised of a bottom edge mode of  
25 the same order or higher than the top edge mode, as shown in Figures 20(a) to 20(d). If  $\epsilon_3 > \epsilon_1$ , then the opposite statement is true.

A careful inspection of the fields associated with the  $ss_b^1$ ,  $sa_b^1$  and  $aa_b^2$  modes reveals that, as the thickness of the film decreases, the mode fields may evolve in a smooth manner similar to that shown in Figures 19(a) to 19(d), but, in addition, a change  
30 or "switch" of the constituent edge modes may also occur. For instance, from Figure 20(c), the  $sa_b^1$  mode is seen to comprise a substrate-metal interface mode having one extremum for a film thickness of 100 nm, while for a thickness of 60 nm the substrate-metal interface mode has three extrema, as shown in Figure 20(d). Since higher-order modes have in general lower phase constants than lower-order modes, this change in  
35 edge modes causes a reduction in the phase constant of the  $sa_b^1$  mode in the neighbourhood of 60 nm, as shown in Figure 18(a). Another change occurs near 40 nm as the corner modes switch from being symmetric-like (as in Figures 20(c) and 20(d)) to being asymmetric-like with respect to the horizontal dimension. This change is again

reflected in the dispersion curve of the  $sa_b^1$  mode as its phase constant is seen to increase with a further decrease in thickness. In general, the changes in the edge and corner modes are consistent with the directions taken by the dispersion curves as the film thickness decreases, thus explaining the oscillations in the curves seen in Figures 18(a) 5 and 18(b).

The only potentially long-ranging mode supported by this structure at the wavelength of analysis is the  $ss_b^1$  mode. As shown in Figures 18(a) and 18(b), the mode has a cutoff thickness near  $t = 22$  nm and although the attenuation drops quickly near this thickness, it should be remembered that the field confinement does so as well.

- 10 Furthermore, the spatial distribution of the main transverse field component related to this mode evolves with decreasing thickness in the manner shown in Figure 20(a) and 20(b), such that near cutoff the spatial distribution has strong extrema along the top and bottom edges. These extrema render the mode less excitable using an end-fire technique, so coupling losses would be higher compared to the fundamental symmetric mode in
- 15 symmetric waveguides. Also, the fact that the mode would be operated near its cutoff thickness implies that very tight tolerances are required in the fabrication of structures. Nevertheless, it should be possible to observe propagation of this mode in a suitable structure using an end-fire experiment.

## 20 VIII. Mode Characteristics and Evolution With Film Thickness: Large Asymmetry

### A. Mode Solutions for a Metal Film Slab Waveguide

The study proceeds with the analysis of structures having a large difference in the dielectric constants of the substrate and superstrate. With respect to Figure 17(a), the relative permittivities of the substrate and superstrate are set to  $\epsilon_{r,1} = 4$  ( $n_1 = 2$ ) and  $\epsilon_{r,3}$  25 = 2.25 ( $n_3 = 1.5$ ), respectively, the width of the metal film is set to  $w = \infty$ , and the dielectric constant of the metal region and the wavelength of analysis are set to the same values as in Section III. The dispersion curves of the  $s_b$  and  $a_b$  modes supported by this structure can be seen in Figures 21(a) and 21(b). Comparing with Figures 18(a) and 18(b), it is observed that the  $s_b$  mode has a larger cutoff thickness in a structure having 30 a large asymmetry than in a structure having similar substrate and superstrate dielectric constants. The results shown were computed using the MoL and are in perfect agreement with those reported in [6].

### B. Modes Supported by a Metal Film of Width $w = 1 \mu m$

35 The structure shown in Figure 17(a) was analyzed using the MoL for  $w = 1 \mu m$  and for the same material parameters and free-space wavelength as those given above for  $w = \infty$ . The dispersion curves of the first six modes supported by the structure are shown in Figures 21(a) and 21(b).

An inspection of the mode fields related to the  $sa_b^0$  and  $aa_b^0$  modes reveals that these modes are again comprised of coupled corner modes with fields localized at the substrate-metal interface. The modes do not change in character as the thickness of the film decreases and a narrowing of the metal film would eventually break the degeneracy  
5 observed in Figures 21(a) and 21(b).

The spatial distribution of the  $E_y$  field component related to the,  $ss_b^0$ ,  $as_b^0$ ,  $sa_b^1$  and  $aa_b^2$  modes is given in Figures 22(a) to 22(d). It is noted from this figure that in all cases the metal-superstrate interface modes are similar: they have fields with no extrema along the interface but rather that are localized near the corners and have either  
10 a symmetric or asymmetric distribution with respect to the  $y$  axis. These corner modes are in fact the lowest order modes supported by the metal-superstrate interface; they have the largest value of phase constant and thus are most likely to couple with edge modes supported by the substrate-metal interface to form a supermode. From Figures 22(a) and 22(b) it is observed that the substrate-metal interface modes comprising  
15 the  $ss_b^0$  and  $as_b^0$  modes are of very high order. This is expected since the substrate dielectric constant is significantly higher than the superstrate dielectric constant and higher order modes have lower values of phase constant. The  $ss_b^0$  and  $as_b^0$  modes shown in Figures 22(a) and 22(b) indeed have fields that are localized along the metal-superstrate interface, while the  $sa_b^1$  and  $aa_b^2$  modes shown in Figures 22(c) and 22(d)  
20 have fields that are localized along the substrate-metal interface.

One effect, caused by increasing the difference between the substrate and superstrate dielectric constants, is that the difference between the orders of the top and bottom edge modes comprising a supermode can increase. This effect can be observed by comparing Figure 19(a) with Figure 22(a). In the former, there is no difference  
25 between the orders of the top and bottom edge modes, while in the latter the difference in the orders is 5. Another effect is that the degree of field localization increases near the interface between the metal and the dielectric of higher permittivity, for all modes that are asymmetric-like with respect to the horizontal dimension. This effect can be seen by comparing the fields related to the  $sa_b^1$  mode shown in Figures 22(c) and 20(c).  
30 A comparison of the fields related to the  $sa_b^0$  and  $aa_b^0$  modes reveals that this effect is present in these modes as well.

From the dispersion curves shown in Figure 21(a), it is apparent that the normalized phase constant of all modes converge with increasing film thickness to normalized phase constants in the neighbourhood of those supported by plasmon-  
35 polariton waves localized along the associated isolated edge. The normalized phase constants of modes having fields localized at the substrate-metal interface, converge with increasing film thickness to normalized phase constants in the neighbourhood of that related to the  $a_b$  mode, while the normalized phase constants of modes having fields

localized along the metal-superstrate interface converge to values near that of the  $s_b$  mode. This behaviour is present though less apparent, in structures where the asymmetry is smaller, such as the one analyzed in Section VII.

Comparing Figures 18(a) and 18(b) with Figures 21(a) and 21(b), it is noted that 5 the dispersion curves of the modes are much smoother when the difference in the substrate and superstrate dielectric constants is large. This is due to the fact that the edge modes comprising the supermodes are less likely to change or switch as they do in a structure having similar substrate and superstrate dielectric constants. Thus modes that start out being symmetric-like with respect to the horizontal dimension remain so as the 10 thickness of the film decreases. The cutoff thickness of the symmetric-like modes also increases as the difference between the substrate and superstrate dielectric constants increases.

It is apparent that introducing a large asymmetry can hamper the ability of the structure to support useful long-ranging modes. Any mode that is long-ranging would 15 likely have fields with numerous extrema along the width of the interface between the metal film and the dielectric of higher permittivity, as shown in Figures 22(a) and 22(b).

## IX. Mode Dispersion with Film Width: Small Asymmetry

An asymmetric structure comprising the same dielectrics as the structures studied 20 in Section VII, but having a metal film of width  $w = 0.5 \mu\text{m}$ , was analyzed at the same free-space wavelength in order to determine the impact of a narrowing film width on the modes supported. The structure was analyzed using the MoL and Figures 23(a) and 23(b) give the dispersion curves obtained for the first few modes supported.

Comparing Figures 23(a) and 23(b) with Figures 18(a) and 18(b) reveals that 25 reducing the width of the film does not cause major changes in the behaviour of the fundamental modes, but does have a major impact on the higher order modes. It is noted that reducing the film width increases the cutoff thickness of the  $ss_b^1$  mode. This higher order mode is symmetric-like with respect to the horizontal dimension, and the cutoff thickness of the symmetric-like modes in general increases as the width of the film 30 decreases due to a reduction in field confinement to the metal film. The  $aa_b^2$  mode was sought but not found for this film width.

It is also noted by comparing Figures 23(a) and 23(b) with Figures 18(a) and 18(b) that the  $sa_b^1$  mode evolves quite differently depending on the width of the film. For a film width of  $w = 1 \mu\text{m}$ , the mode follows the general behaviour of an 35 asymmetric-like mode whereas, for a film width of  $w = 0.5 \mu\text{m}$ , the mode evolves as a symmetric-like mode, and has a cutoff thickness near  $t = 27 \text{ nm}$ . When the film is wide, it becomes possible for numerous higher order edge modes (having similar values of phase constant) to be supported by the substrate-metal or metal-superstrate interfaces,

so edge modes comprising a supermode are likely to change or switch as the thickness of the film is reduced, as shown in Figures 20(c) and 20(d). For a narrow metal film, some of the higher order edge modes may be cutoff, thus rendering changes in edge modes impossible. In such a case, the supermode may be forced to evolve in a smooth manner with decreasing film thickness. A close inspection of the mode fields related to the  $sa_b^1$  mode for a film width of  $w = 0.5 \mu\text{m}$  reveals that there are no changes to the edge modes as the thickness decreases; rather the mode evolves smoothly from its field distribution at a large thickness (similar to that shown in Figure 20(c)) to having a symmetric-like distribution with only one extremum along the top and bottom edges of the film. A change in behaviour due to a change in the width of the metal film was observed only for the  $sa_b^1$  mode in this study, but such changes are in general not limited to this mode.

The  $sa_b^1$  and  $ss_b^1$  modes could be made to propagate over useful distances in this structure, if they are excited near their cutoff thicknesses. However, the difficulties outlined in Section VII B regarding the excitation of modes near cutoff also hold here.

#### X. Evolution of the $ss_b^0$ and $sa_b^1$ Modes With Structure Asymmetry

The  $ss_b^0$  and  $sa_b^1$  modes are of practical interest. The  $ss_b^0$  mode is the main long-ranging mode supported by symmetric finite-width metal film structures, and, as demonstrated in the previous section, the  $sa_b^1$  mode can be the main long-ranging mode supported by asymmetric finite-width structures. In metal films of the right thickness, they are also the modes that are the most suitable to excitation in an end-fire arrangement.

Structures comprising a substrate dielectric having  $n_1 = 2$ , of a metal film of width  $w = 0.5 \mu\text{m}$ , and of various superstrate dielectrics having  $n_3 = 2, 1.99, 1.95$  and 1.9 were analyzed at the same free-space wavelength as in Section VII. The equivalent permittivity of the metal film was also set to the same value as in Section VII. These analyses were performed in order to investigate the effects on the propagation characteristics of the  $ss_b^0$  and  $sa_b^1$  modes caused by a slight decrease in the superstrate permittivity relative to the substrate permittivity. Figures 24(a) and 24(b) show the dispersion curves with film thickness, obtained for these modes in the four structures of interest.

As seen in Figure 24(a) and its inset, the dispersion curves of the modes intersect at a certain film thickness only for the symmetric case ( $n_3 = n_1$ ). As soon as some degree of asymmetry exists, the curves no longer intersect, though they may come quite close to each other if the asymmetry is small, as seen in the case of  $n_3 = 1.99$ . As the degree of asymmetry increases, the separation between the curves increases.

The evolution with film thickness of the  $sa_b^1$  mode is shown in Figures 25(a) to 25(d) for the case  $n_3 = 1.99$  and for thicknesses about  $t = 59$  nm (near the maximum in its phase dispersion curve). The evolution of this mode for the cases  $n_3 = 1.95$  and 1.9 is similar to that shown. The evolution with film thickness of the  $ss_b^0$  mode is 5 similar in these structures to the evolution shown in Figures 19(a) to 19(d) for the case  $w = 1 \mu\text{m}$  and  $n_3 = 1.9$ . Comparing Figures 25(a) to 25(d) and Figures 19(a) to 19(d), reveals that the modes "swap" character near  $t = 59$  nm. For film thicknesses sufficiently above this value, the modes exhibit their defining character as shown in Figures 19(a) and 25(a), but for film thicknesses below it, each mode exhibits the other's 10 character, as shown in Figures 19(d) and 25(d). This character swap is present for the three cases of asymmetry considered here ( $n_3 = 1.99, 1.95$  and 1.9) and explains the behaviour of the dispersion curves shown in Figures 24(a) and 24(b).

From Figures 24(a) and 24(b), it is noted that a cutoff thickness exists for the long-ranging mode as soon as an asymmetry is present in the structure. It is also 15 observed that the cutoff thickness increases with increasing asymmetry. In the case of  $n_3 = 1.99$ , the cutoff thickness of the mode is near  $t = 12$  nm, while for  $n_3 = 1.9$  the cutoff thickness is near  $t = 27$  nm. As the width of the metal film  $w$  increases, the cutoff thickness of the  $sa_b^1$  mode decreases as long as the mode remains long-ranging (recall that the character of this mode may also change with film width such that its 20 behaviour is similar to the  $a_b$  mode in the corresponding slab structure, as shown in Figures 18(a) and 18(b)). Also, it is clear from Figures 23(a) and 23(b) that the cutoff thickness of the  $sa_b^1$  mode is greater than the cutoff thickness of the  $s_b$  mode supported by the corresponding slab structure. These results imply that the long-ranging mode supported by a thin narrow metal film is more sensitive to differences in the superstrate 25 and substrate permittivities than the  $s_b$  mode supported by the corresponding slab structure. This is reasonable in light of the fact that, in finite-width structures, the mode fields tunnel through the metal as in slab structures, but, in addition, the fields also wrap around the metal film.

Figure 24(b) shows that near cutoff, the attenuation of the  $sa_b^1$  mode supported 30 by an asymmetric structure drops much more rapidly than the attenuation of the  $ss_b^0$  mode supported by a symmetric structure. Thus, a means for range extension, similar to that observed in asymmetric slab structures [7], exists for metal films of finite width, though the difficulties related to the excitation of a mode near its cutoff thickness, as described in Section VII B, also apply here.

35 Figures 26(a) to 26(d) show contour plots of  $\Re\{S_z\}$  associated with the long-ranging modes for the four cases of superstrate permittivity considered.  $S_z$  is the z-directed component of the Poynting vector and its spatial distribution is computed from the spatial distribution of the mode fields using:

$$S_z = (E_x H_y^* - E_y H_x^*)/2 \quad (6)$$

where  $H_{x,y}^*$  denotes the complex conjugate of  $H_{x,y}$ . Figure 26(a) shows the contour plot associated with the  $ss_b^0$  mode supported by a symmetric structure ( $n_3 = n_1 = 2$ ) of thickness  $t = 20$  nm. Figures 26(b),(c) and (d) show contours associated with the  $sa_b^1$  mode for the three cases of structure asymmetry considered. The contour plots shown in Figures 26(b),(c) and (d) are computed for film thicknesses slightly above cutoff, representative of the thicknesses that would be used to observe these long-ranging modes experimentally. From those figures, it is noted that the contour plots become increasingly distorted and the fields increasingly localized at the metal-superstrate interface as the degree of asymmetry in the structure increases. It is also apparent by comparing Figures 26(a) and 26(d), that in an end-fire experiment, less power should be coupled into the  $sa_b^1$  mode supported by the asymmetric structure with  $n_3 = 1.9$ , compared to the  $ss_b^0$  mode supported by the symmetric structure. End-fire coupling losses seem to increase with increasing structure asymmetry.

The high sensitivity of the long-ranging mode supported by thin metal films of finite width, to structure asymmetry, is potentially useful. A small induced asymmetry (created via an electro-optic effect present in the dielectrics say) can evidently effect a large change in the propagation characteristics of the long-ranging mode. From Figures 24(a) and 24(b), it is apparent that a difference between the substrate and superstrate refractive indices as small as  $n_1 - n_3 = \Delta n = 0.01$  is sufficient to create an asymmetric structure where the long-ranging mode has a cutoff thickness of about  $t = 12$  nm. From Figure 24(a), a slightly larger difference of  $\Delta n = 0.05$  changes the normalized phase constant of the long-ranging mode by  $\Delta(\beta/\beta_0) \approx 0.025$  for a metal film thickness of  $t = 20$  nm. Both of these effects are potentially useful.

Asymmetric structures having superstrate dielectric constants that are slightly greater than that of the substrate were also analyzed. The substrate dielectric constant was set to  $n_1 = 2$  and superstrate dielectrics having  $n_3 = 2.01, 2.05$  and  $2.1$  were considered for the same metal, film width and operating wavelength. The results are similar to those presented in Figures 24 through 26 and the cut-off thicknesses are near those shown in Figure 24(b). Though the results are not identical, there is no major difference between the behaviour of the  $ss_b^0$  and  $sa_b^1$  modes whether  $\epsilon_1 > \epsilon_3$  or  $\epsilon_1 < \epsilon_3$  as long as the permittivities are similar.

### 35 XI. Conclusion

The purely bound optical modes supported by thin lossy metal films of finite width, embedded in an "infinite" homogeneous dielectric have been characterized and described. The modes supported by these symmetric structures are divided into four

families depending on the symmetry of their mode fields and none of the modes are TM in nature (as they are in the metal film slab waveguide). In addition to the four fundamental modes that exist, numerous higher order modes are supported as well. A proposed mode nomenclature suitable for identifying them has been discussed. The dispersion of the modes with film thickness has been assessed and the behaviour in general terms found to be consistent with that of the purely bound modes supported by the metal film slab waveguide. In addition, it has been found that one of the fundamental modes and some higher order modes have cut-off thicknesses. Mode dispersion with film width has also been investigated and it has been determined that the higher order modes have a cut-off width, below which they are no longer propagated. The effect of varying the background permittivity on the modes has been investigated as well, and the general behaviour found to be consistent with that of the modes supported by a metal film slab waveguide. In addition it was determined that the cut-off width of the higher order modes decreases with decreasing background permittivity and that all cut-off thicknesses are increased.

One of the fundamental modes supported by the symmetric structures, the  $ss_b^0$  mode, exhibits very interesting characteristics and is potentially quite useful. This mode evolves with decreasing film thickness towards the TEM wave supported by the background, (an evolution similar to that exhibited by the  $s_b$  mode in metal film slab waveguides), its losses and phase constant tending asymptotically towards those of the TEM wave. In addition, it has been found that decreasing the film width can reduce the losses well below those of the  $s_b$  mode supported by the corresponding metal film slab waveguide. Reducing the background permittivity further reduces the losses. However, a reduction in losses is always accompanied by a reduction in field confinement to the waveguide core, which means that both of these parameters must be traded-off one against the other. Furthermore, carefully selecting the film's thickness and width can make the  $ss_b^0$  mode the only long-ranging mode supported. It has also been demonstrated that mode power attenuation values in the range of 10 to 0.1 dB/cm are achievable at optical communications wavelengths, with even lower values possible. Finally, evolved into its most useful form, the  $ss_b^0$  mode has a field distribution that renders it excitable using end-fire techniques.

The existence of the  $ss_b^0$  mode in a symmetric structure, as well as its interesting characteristics, makes the finite-width metal film waveguide attractive for applications requiring short propagation distances. The waveguide offers two-dimensional field confinement in the transverse plane, rendering it useful as the basis of an integrated optics technology. Interconnects, power splitters, power couplers and interferometers could be built using the guides. Finally, the structures are quite simple and so should be inexpensive to fabricate.

The long-ranging modes supported by asymmetric structures of finite width have a rapidly diminishing attenuation near their cutoff thickness (like asymmetric slab structures). The rate of decrease of the attenuation with decreasing thickness near cutoff is greater than the rate related to the  $ss_b^0$  mode in symmetric structures. However field confinement also diminishes rapidly near cutoff, implying that the structures ought to be fabricated to very tight tolerances and that all metal-dielectric interfaces should be of the highest quality. It has also been found that decreasing the width of the film increases the cutoff thickness of the main long-ranging mode. Below this cutoff thickness, no purely bound long-ranging mode exists. The long-ranging modes supported by metal films of finite-width are thus more sensitive to the asymmetry in the structure as compared to the  $s_b$  mode supported by similar slab waveguides. This is a potentially useful result in that a small induced change in substrate or superstrate refractive index can have a greater impact on the long-ranging mode supported by a finite-width structure as compared to a similar slab waveguide.

Parts of the foregoing theoretical discussion have been published by the inventor in references [13] [20], [44] and [45].

#### SPECIFIC EMBODIMENTS AND EXAMPLES OF APPLICATION

Examples of practical waveguide structures, and integrated optics devices which can be implemented using the invention, will now be described with reference also to Figures 27 to 42. Unless otherwise stated, where a waveguide structure is shown, it will have a general construction similar to that shown in Figures 1(a) and 1(b) or that shown in Figures 17(a) and 17(b).

The waveguide structure 100 shown in Figures 1(a) and 1(b) comprises a strip of finite thickness  $t$  and width  $w$  of a first material having a high free (or almost free) charge carrier density, surrounded by a second material which has a very low free carrier density. The strip material can be a metal or a highly doped semiconductor and the background material can be a dielectric.

Suitable materials for the strip include (but are not limited to) gold, silver, copper, aluminium and highly n- or p-doped GaAs, InP or Si, while suitable materials for the surrounding material include (but are not limited to) glass, quartz, polymer and undoped or very lightly doped GaAs, InP or Si. Particularly suitable combinations of materials include Au for the strip and SiO<sub>2</sub> for the surrounding material.

The thickness  $t$  and the width  $w$  of the strip are selected such that the waveguide supports a long-ranging plasmon-polariton mode at the free-space operating wavelength of interest. Suitable dimensions for Au/SiO<sub>2</sub> waveguides at an operating free-space wavelength of 1550 nm are about 10 to 30 nm for the thickness  $t$  and about 2 to 12  $\mu\text{m}$  for the width  $w$ ; a thickness of 20 nm and a width of 4  $\mu\text{m}$  are good dimensions.

Figures 51 and 52 illustrate mode power attenuation for waveguides constructed from strips of gold (Au) and aluminium (Al), respectively, each embedded in silicon dioxide ( $\text{SiO}_2$ ) for various widths and thicknesses of the metal film. Analyses were carried out with an optical free space wavelength of 1.55 microns. The curves show that

5 very low attenuation values can be obtained with metal strips of practical dimensions. Generally, the attenuation using the gold strip was about one half of that obtained with aluminium strip having similar dimensions. In both cases, thicknesses ranging from about 15 to 30 nm and widths ranging from about 1 to 8 microns gave particularly good results.

10 Unless otherwise stated when structure dimensions are mentioned from this point onward, they refer to the Au/ $\text{SiO}_2$  material combination at an operating optical free-space wavelength of 1550 nm. Similar dimensions are needed for most material combinations.

The plasmon-polariton field may be excited by optical radiation coupled to the strip in an end-fire manner from a fiber butt-coupled to the input of the waveguide. The

15 output of the waveguide can also be butt-coupled to a fibre. Alternatively, the waveguide could be excited at an intermediate position by an alternative means, for example using the so-called attenuated total reflection method (ATR).

The length  $l$  shown in Figure 1(b) is arbitrary and will be selected to implement a desired interconnection.

20 It has been demonstrated that a straight waveguide 100 with the dimensions set out above is polarisation sensitive. The plasmon-polariton wave is highly linearly polarised in the vertical direction, i.e. perpendicular to the plane of the strip. Hence, it may serve as a polarisation filter, whereby substantially only a vertical polarised mode (aligned along the y-axis as defined in Figure 1(a)) of the incident light is guided.

25 Figure 27 shows a transition waveguide section 102 having stepped sides which can be used to interconnect two sections of waveguide having different widths. The larger width can be used to more effectively couple the waveguide to the input/output fibres. The reduced width helps to reduce the insertion loss of the waveguide. Typical widths are about  $W_2=10 \mu\text{m}$  to couple to single mode fibre and  $W_1=4 \mu\text{m}$  for the

30 waveguide width. Any symmetry of the structure shown can be used.

Figure 28 shows an angled section 104 which can be used as an interconnect. Its dimensions,  $W_1$ ,  $W_2$  and  $l$  and the angles  $\phi_1$  and  $\phi_2$ , are adjusted for a particular application as needed. Usually the angles are kept small, in the range of 1 to 15 degrees and the input and output widths are usually similar, about  $4 \mu\text{m}$ . Although the sides of

35 the angled section 104 shown in Figure 28 are tapered, they could be parallel. It should also be appreciated that the angle of the inclination could be reversed, i.e. the device could be symmetrical about the bottom right hand corner shown in Figure 28 or transposed about that axis if not symmetrical about it.

Figure 29 shows a tapered waveguide section 106, which can be used to interconnect two waveguides of different widths. The length of the taper is usually adjusted such that the angles are small, usually in the range of 1 to 15 degrees. The taper angles at the two sides are not necessarily the same. Such a configuration might 5 be used as an input port, perhaps as an alternative to the layout shown in Figure 27, or as part of another device, such as a power splitter. Any symmetry of the structure shown can be used.

Figure 30 illustrates an alternative transition waveguide section 130 which has curved sides, rather than straight as in the trapezoidal transition section disclosed in 10 Figure 29. In Figure 30, the curved sides are shown as sections of circles of radius  $R_1$  and  $R_2$ , subtending angles  $\phi_1$  and  $\phi_2$  respectively, but it should be appreciated that various functions can be implemented, such as exponential or parabolic, such that the input and output reflections are minimised.

Figure 31 shows a curved waveguide section 108 which can be used to redirect 15 the plasmon-polariton wave. The angle  $\phi$  of the bend can be in the range of 0 to 360 degrees and the bending radius  $R$  can be in the range of a few microns to a few centimetres. For a 45-degree bend, a radius of 0.5 to 2 cm is appropriate. The critical dimensions are the radius  $R$  and the positions of the input and output straight sections 100. Although the device will work, and the structure 108 will convey the plasmon- 20 polariton wave around the bend, there is leakage out of the bend (from the exterior curve) and also reflection back in the direction from which the wave came. Reduced radiation and reflection is obtained when the input and output waveguides 100 are offset outwards relative to the ends of the bend. The reason for this is that the straight waveguide sections 100 have an optical field extremum that peaks along the longitudinal 25 centre line, and then decays towards the edges. In the bend, the extremum of the optical field distribution shifts towards the exterior of the curve. This results in increased radiation from the external edge of the curve and increased reflection back to the input waveguide 100 due to a mismatch in the field distributions. Offsetting the input and output waveguides 100 towards the outside of the curve aligns the extrema of their 30 optical fields more closely with that of the optical field in the curved section 108, which helps to reduce, even minimise both the radiation and the reflection. The tighter the radius  $R$ , the greater the radiation from the exterior of the curve, so the offset  $O_1$  is related to the radius  $R$  and the optimum values would have to be determined according to the specific application.

35 It should also be noted that it is not necessary to connect the input and output waveguides 100 directly to the curve. As shown in Figure 31, it is possible to have a short spacing  $d_1$  between the end of the input waveguide 100 and the adjacent end of the curved section 108. Generally speaking, that spacing  $d_1$  should be minimised, even zero,

and probably no more than a few optical wavelengths. A similar offset  $O_2$  and spacing  $d_2$  could be provided between the bend 108 and the output straight waveguide 100.

Although Figure 31 shows no gradual transition between the straight waveguides 100 at the input and output and the ends of the curved section 108, it is envisaged that,  
5 in practice, a more gradual offset could be provided so as to reduce edge effects at the corners.

Figure 32 shows a two-way power splitter 110 formed from a trapezoidal section 106 with a straight section 100 coupled to its narrower end 112 and two angled sections 104 coupled side-by-side to its wider end 114. The distances between the input  
10 waveguide 100 and the narrower end 112 of the tapered section 106, and between the output waveguides 104 and the wider end 114 of the tapered section 106,  $d_1$ ,  $d_2$  and  $d_3$ , respectively, should be minimised. The angle between the output waveguides 104 is usually in the range of 0.5 to 10 degrees and their widths are usually similar. The offsets  $S_1$  and  $S_2$  between the output waveguides and the longitudinal centre line of the  
15 trapezoidal section 106 preferably are set to zero, but could be non-zero, if desired, and vary in size. Ideally, however, the output sections 104 should together be equal in width to the wider end 114.

Offset  $S_1$  need not be equal to offset  $S_2$  but it is preferable that both are set to zero. The widths of the output sections 104 can be adjusted to vary the ratio of the  
20 output powers. The dimensions of the centre tapered section 106 are usually adjusted to minimise input and output reflections and radiation losses in the region between the output sections 104.

It should also be noted that the centre tapered section 106 could have angles that vary according to application and need not be symmetrical.

25 It is envisaged that the tapered section 106 could be replaced by a rectangular transition section having a width broader than the width of the input waveguide 100 so that the transition section favoured multimode propagation causing constructive/destructive interference patterns throughout its length. The length could be selected so that, at the output end of the rectangular transition section, the constructive  
30 portions of the interference pattern would be coupled into the different waveguides establishing, in effect, a 1-to-N power split. Such a splitter then would be termed a multimode interferometer-based power divider.

It should be appreciated that the device shown in Figure 32 could also be used as a combiner. In this usage, the light would be injected into the waveguide sections 104  
35 and combined by the tapered centre section 106 to form the output wave which would emerge from the straight waveguide section 100.

In either the Y splitter or the interferometer-based power divider, the number of arms or limbs 104 at the output could be far more than the two that are shown in Figure 32.

It is also feasible to have a plurality of input waveguides. This would enable an 5 NxN divider to be constructed. The dimensions of the transition section 106 then would be controlled according to the type of splitting/combining required.

As shown in Figure 33, an angled waveguide section 104 may be used to form an intersection between two straight waveguide section 100, with the dimensions adjusted for the particular application. It should be noted that, as shown in Figure 32, the two 10 straight sections 100 are offset laterally away from each other by the distances  $O_1$  and  $O_2$ , respectively, which would be selected to optimise the couplings by reducing radiation and reflection losses, in the manner discussed with reference to Figure 31. The angle of the trapezoidal section 104 will be a factor in determining the best values for the offsets  $O_1$  and  $O_2$ . The sections 100 and 104 need not be connected directly together, 15 but could be spaced by the distances  $d_1$  and  $d_2$  and/or coupled by a suitable transition piece that would make the junction more gradual (i.e., the change of direction would be more gradual).

The embodiments of Figures 31 and 32 illustrate a general principle of aligning optical fields, conveniently by offsets, wherever there is a transition or change of 20 direction of the optical wave and an inclination relative to the original path, which can cause radiation and reflection if field extrema are misaligned. Such offsets would be applied whether the direction-changing sections were straight or curved.

As illustrated in Figure 34, a power divider 116 can also be implemented using a pair of concatenated curved sections 108 instead of each of the angled sections 104 in 25 the splitter 110 shown in Figure 32. As shown in Figure 34, in each pair, the curved section nearest to the wider end 114 of the tapered section 106 curves outwards from the longitudinal centre line of the tapered section 106 while the other curved section curves oppositely so that they form an "S" bend. Also, the curved sections in each pair are offset by distance  $O_1$  or  $O_2$  one relative to the other for the reasons discussed with 30 respect to the bend 108 shown in Figure 31. Other observations made regarding the power divider and the curved section disclosed in Figures 31 and 32 respectively, also hold in this case.

Figure 35 illustrates a Mach-Zehnder interferometer 118 created by interconnecting two power splitters 110 as disclosed in Figure 32. Of course, either or 35 both of them could be replaced by the power splitter 116 shown in Figure 34. Light injected into one of the ports, i.e. the straight section 100 of one power splitter 110/116, is split into equal amplitude and phase components that travel along the angled arms 104

of the splitter, are coupled by straight sections 100 into the corresponding arms of the other splitter, and then are recombined to form the output wave.

If the insertion phase along one or both arms of the device is modified, then destructive interference between the re-combined waves can be induced. This induced 5 destructive interference is the basis of a device that can be used to modulate the intensity of an input optical wave. The lengths of the arms 100 are usually adjusted such that the phase difference in the re-combined waves is 180 degrees for a particular relative change in insertion phase per unit length along the arms. The structure will thus be optically long if the mechanism used to modify the per unit length insertion phase is weak (or 10 optically short if the mechanism is strong).

Figure 36(a) illustrates a modulator 120 based on the Mach-Zehnder 118 disclosed in Figure 35. As illustrated also in Figure 36(b), parallel plate electrodes 122 and 124 are disposed above and below, respectively, each of the strips 100 which interconnects two angled sections 104, and spaced from it, by the dielectric material, at a distance 15 large enough that optical coupling to the electrodes is negligible. The electrodes are connected in common to one terminal of a voltage source 126, and the intervening strip 100 is connected using a minimally invasive contact to the other terminal. Variation of the voltage V applied by source 126 effects the modulating action. According to the plasma model for the strip 100, a change in the carrier density of the latter (due to 20 charging  $+2Q$  or  $-2Q$ ) causes a change in its permittivity, which in turn causes a change in the insertion phase of the arm. (The change induced in the permittivity is described by the plasma model representing the guiding strip 100 at the operating wavelength of interest. Such model is well known to those of ordinary skill in the art and so will not be described further herein. For more information the reader is directed to reference 25 [21], for example.) This change is sufficient to induce destructive interference when the waves in both arms re-combine at the output combiner.

Figure 36(c) illustrates an alternative connection arrangement in which the two plate electrodes 122 and 124 are connected to respective ones of the terminals of the voltage source 126. In this case, the dielectric material used as the background of the 30 waveguide is electro-optic ( $\text{LiNbO}_3$ , a polymer,...). In this instance, the applied voltage V effects a change in the permittivity of the background dielectric, thus changing the insertion phase along the arm. This change is sufficient to induce destructive interference when the waves in both arms re-combine at the output combiner.

It will be noted that, in Figure 36(a), one voltage source supplies the voltage  $V_1$  35 while the other supplies the voltage  $V_2$ .  $V_1$  and  $V_2$  may or may not be equal.

For both cases described above, it is possible to apply voltages in opposite polarity to both arms of the structure. This effects an increase in the insertion phase of one arm and a decrease in the insertion phase of the other arm of the Mach-Zehnder (or

vice versa), thus reducing the magnitude of the voltage or the length of the structure required to achieve a desirable degree of destructive interference at the output.

Also, it is possible to provide electrodes 122 and 124 and a source 126 for only one of the intervening strips 100 in order to provide the required interference.

5 It should be appreciated that other electrode structures could be used to apply the necessary voltages. For example, the electrodes 122 and 124 could be coplanar with the intervening strip 100, one on each side of it. By carefully laying out the electrodes as a microwave waveguide, a high frequency modulator (capable of modulation rates in excess of 10 Gbit/s) can be achieved.

10 Figure 37 illustrates an alternative implementation of a Mach-Zehnder 128 which has the same set of waveguides as that shown in Figure 35 but which makes use of magnetic fields  $B$  applied to either or both of the middle straight section arms to induce a change in the permittivity tensor describing the strips. (The change induced in the tensor is described by the plasma model representing the guiding strip at the operating 15 wavelength of interest. Such model is well known to those of ordinary skill in the art and so will not be described further herein. For more information the reader is directed to reference [21], for example.) The change induced in the permittivity tensor will induce a change in the insertion phase of either or both arms thus inducing a relative phase difference between the light passing in the arms and generating destructive 20 interference when the waves re-combine at the output combiner. Modulating the magnetic field thus modulates the intensity of the light transmitted through the device. The magnetic field  $B$  can be made to originate from current-carrying wires or coils disposed around the arms 100 in such a manner as to create the magnetic field in the desired orientation and intensity in the optical waveguides. The magnetic field may have 25 one or all of the orientations shown,  $B_x$ ,  $B_y$  or  $B_z$  or their opposites. The wires or coils could be fabricated using plated via holes and printed lines or other conductors in known manner. Alternatively, the field could be provided by an external source, such as a solenoid or toroid having poles on one or both sides of the strip.

Figure 38 illustrates a periodic waveguide structure 132 comprising a series of 30 unit cells 134, where each cell 134 comprises two rectangular waveguides 100 and 100' having different lengths  $l_1$  and  $l_2$  and widths  $w_1$  and  $w_2$ , respectively. The dimensions of the waveguides in each unit cell 134, the spacing  $d_1$  therebetween, the number of unit cells, and the spacings  $d_2$  between cells are adjusted such that reflection occurs at a desired operating wavelength or over a desired operating bandwidth for an optical signal 35 propagating along the grating axis, i.e. the longitudinal axis of the array of cells 134. The period of the periodic structure, i.e. the length of each unit cell,  $l_1 + l_2 + d_1 + d_2$ , can be made optically long, such that a long-period periodic structure is obtained. The dimensions of the elements 100, 100' in each unit cell 134 can also be made to change

along the direction of the periodic structure in order to implement a prescribed transfer function (like in a chirped periodic structure).

It should be noted that the waveguides 100, 100' in each cell need not be rectangular, but a variety of other shapes could be used. For example, Figure 39 5 illustrates a portion, specifically two unit cells 138 only, of an alternative periodic structure 136 in which each unit cell 138 comprises two of the trapezoidal waveguide sections 106, 106' like that described with reference to Figure 30, with their wider edges opposed.

As another alternative, the trapezoidal waveguides 106/106' could be replaced by 10 the transition sections 130, shown in Figure 30, with or without spacings  $d_1$  and  $d_2$ , to form a periodic structure having sinusoidally-varying sides. It should be noted that these periodic structures are merely examples and not intended to provide an exhaustive detailing of all possibilities; various other periodic structures could be formed from unit cells comprised of all sorts of different shapes and sizes of elements.

15 It should be noted that voltages can be applied to some or all of the strips in order to establish charges on the strips of the unit cells, which would change their permittivity and thus vary the optical transfer function of the periodic structures. If the dielectric material surrounding the strip is electro-optic, then the applied voltages would also change the permittivity of the dielectric, which also contributes to changing of the optical 20 transfer function of the periodic structure.

Photonic bandgap structures can be created by placing two-dimensional arrays of unit cells (comprised of strips of various shapes and sizes) over numerous planes separated by dielectric material. The size and shape of the strips are determined such that stop bands in the optical spectrum appear at desired spectral locations.

25 Figure 40(a) illustrates an edge coupler 139 created by placing two strips 100" parallel to each other and in close proximity over a certain length. The separation  $S_c$  between the strips 100" could be from 1  $\mu\text{m}$  (or less) to 20  $\mu\text{m}$  and the coupling length  $L_c$  could be in the range of a few microns to a few dozen millimeters depending on the separation  $S_c$ , width and thickness of the strips 100", the materials used, the operating 30 wavelength, and the level of coupling desired. Such a positioning of the strips 100" is termed "edge coupling".

The gaps between the input and output of the waveguide sections shown would ideally be set to zero and a lateral offset provided between sections where a change of direction is involved. Curved sections could be used instead of the sections 104, 100 and 35 100" shown in Figure 40(a).

Although only two strips 100" are shown in the coupled section, it should be understood that more than two strips can be coupled together to create an NxN coupler.

As illustrated in Figure 40(b) a voltage can be applied to the two edge-coupled sections 100" via minimally invasive electrical contacts. Figure 40(b) shows a voltage source 126 connected directly to the sections 100" but, if the sections 100, 104 and 100" in each arm are connected together electrically, the source 126 could be connected to one 5 of the other sections in the same arm. Applying a voltage in such a manner charges the arms of the coupler, which, according to the plasma model for the waveguide, changes its permittivity. If, in addition, the dielectric material placed between the two waveguides 100 is electro-optic, then a change in the background permittivity will also be effected as a result of the applied voltage. The first effect is sufficient to change the 10 coupling characteristics of the structure but, if an electro-optic dielectric is also used, as suggested, then both effects will be present, allowing the coupling characteristics to be modified by applying a lower voltage.

Figures 41(a) and 41(b) illustrate coupled waveguides similar to those shown in Figure 40(a) but placed on separate layers in a substrate having several layers 140/1, 15 140/2 and 140/3. The strips could be placed one directly above the other with a thin region of dielectric of thickness d placed between them. Such positioning of the strips is termed "broadside coupling". The coupled guides can also be offset from broadside a distance S<sub>c</sub>, as shown in Figures 41(a) and 41(b). The strips could be separated by d=1 μm (or less) to 20 μm, the coupling length could be in the range of a few microns to a 20 few dozens millimeters and the separation S<sub>c</sub> could be in the range of -20 to +20 μm, depending on the width and thickness of the strips, the materials used and the level of coupling desired. (The negative indicates an overlap condition)

As before, curved sections could be used instead of the straight and angled sections shown in Figure 41(a).

25 Gaps can be introduced longitudinally between the segments of strip if desired and a lateral offset between the straight and angled (or curved) sections could be introduced.

Though only two strips are shown in the coupled section, it should be understood that a plurality of strips can be coupled together on a layer and/or over many layers to create an NxN coupler.

30 As shown in Figure 41(b), a voltage source 126 could be connected directly or indirectly to the middle (coupled) sections 100" in a similar manner to that shown in Figure 40(b).

As illustrated in Figure 42, an intersection 142 can be created by connecting together respective ends of four of the angled waveguide sections 104, their distal ends 35 providing input and output ports for the device. When light is applied to one of the ports, a prescribed ratio of optical power emerges from the output ports at the opposite side of the intersection. The angles  $\phi_1 \dots \phi_4$  can be set such that optical power input into

one of the ports emerges from the port directly opposite, with negligible power transmitted out of the other ports. Any symmetry of the structure shown is appropriate.

Various other modifications and substitutions are possible without departing from the scope of the present invention. For example, although the waveguide structure 5 shown in Figures 1(a) and 1(b), and implicitly those shown in other Figures, have a single homogeneous dielectric surrounding a thin metal film, it would be possible to sandwich the metal film between two slabs of different dielectric material; or at the junction between four slabs of different dielectric material. Moreover, the multilayer dielectric material(s) illustrated in Figure 41(a) could be used for other devices too.

10 Also, the thin metal film could be replaced by some other conductive material or a highly n- or p-doped semiconductor. It is also envisaged that the conductive film, whether metal or other material, could be multi-layered.

#### Specific embodiments of modulation and switching devices

15 Modulation and switching devices will now be described which make use of the fact that an asymmetry induced in optical waveguiding structures having as a guiding element a thin narrow metal film may inhibit propagation of the main long-ranging purely bound plasmon-polariton mode supported.

The asymmetry in the structure can be in the dielectrics surrounding the metal 20 film. In this case the permittivity, permeability or the dimensions of the dielectrics surrounding the strip can be different. A noteworthy case is where the dielectrics above and below the metal strip have different permittivities, in a manner similar to that shown in Figure 17(a).

A dielectric material exhibiting an electro-optic, magneto-optic, thermo-optic, or 25 piezo-optic effect can be used as the surrounding dielectric medium. The modulation and switching devices make use of an external stimulus to induce or enhance the asymmetry in the dielectrics of the structure.

Figures 43(a) and 43(b) depict an electro-optic modulator comprising two metal strips 110 and 112 surrounded by a dielectric 114 exhibiting an electro-optic effect. Such 30 a dielectric has a permittivity that varies with the strength of an applied electric field. The effect can be first order in the electric field, in which case it is termed the Pockels effect, or second order in the electric field (Kerr effect), or higher order. Figure 43(a) shows the structure in cross-sectional view and Figure 43(b) shows the structure in top view. The lower metal strip 110 and the surrounding dielectric 114 form the optical 35 waveguide. The lower metal strip 110 is dimensioned such that a purely bound long-ranging plasmon-polariton wave is guided by the structure at the optical wavelength of interest. Since the "guiding" lower metal strip 110 comprises a metal, it is also used as an electrode and is connected to a voltage source 116 via a minimally invasive electrical

contact 118 as shown. The second metal strip 112 is positioned above the lower metal strip 110 at a distance large enough that optical coupling between the strips is negligible. It is noted that the second strip also be placed below the waveguiding strip instead of above. The second strip acts as a second electrode.

5       The intensity of the optical signal output from the waveguide can be varied or modulated by varying the intensity of the voltage  $V$  applied by the source 116. When no voltage is applied, the waveguiding structure is symmetrical and supports a plasmon-polariton wave. When a voltage is applied, an asymmetry in the waveguiding structure is induced via the electro-optic effect present in the dielectric 114, and the propagation 10 of the plasmon-polariton wave is inhibited. The degree of asymmetry induced may be large enough to completely cut-off the main purely bound long-ranging mode, thus enabling a very high modulation depth to be achieved. By carefully laying out the electrodes as a microwave waveguide, a high frequency modulator (capable of modulation rates in excess of 10 G bit/s) can be achieved.

15     Figures 44(a) and 44(b) show an alternative design for an electro-optic modulator which is similar to that shown in Figure 43(a) but comprises electrodes 112A and 112B placed above and below, respectively, the metal film 110 of the optical waveguide at such a distance that optical coupling between the strips is negligible. Figure 44(a) shows the structure in cross-sectional view and Figure 44(b) shows the structure in top view.

20     A first voltage source 116A connected to the metal film 110 and the upper electrode 112A applies a first voltage  $V_1$  between them. A second voltage source 116B connected to metal film 110 and lower electrode 112B applies a voltage  $V_2$  between them.. The voltages  $V_1$  and  $V_2$ , which are variable, produce electric fields  $E_1$  and  $E_2$  in portions 114A and 114B of the dielectric material. The dielectric material used exhibits an 25 electro-optic effect. The waveguide structure shown in Figure 44(c) is similar in construction to that shown in Figure 44(a) but only one voltage source 116C is used. The positive terminal (+) of the voltage source 116C is shown connected to metal film 110 while its negative terminal (-) is shown connected to both the upper electrode 112A and the lower electrode 112B. With this configuration, the electric fields  $E_1$  and  $E_2$

30     produced in the dielectric portions 114A and 114B, respectively, are in opposite directions. Thus, whereas, in the waveguide structure of Figure 44(a), selecting appropriate values for the voltages  $V_1$  and  $V_2$  induces the desired asymmetry in the waveguide structure of Figure 44(c), the asymmetry is induced by the relative direction of the electric field above and below the waveguiding strip 110, since the voltage  $V$  35 applied to the electrodes 112A and 112B produces electric fields acting in opposite directions in the portions 114A and 114B of the dielectric material.

The structures shown in Figures 44(a),(b) and (c) can operate to very high frequencies since a microwave transmission line (a stripline) is in effect created by the three metals.

Figure 45 shows in cross-sectional view yet another design for an electro-optic modulator. In this case, the metal film 110 is embedded in the middle of dielectric material 114 with first portion 114D above it and second portion 114E below it. Electrodes 112D and 112E are placed opposite lateral along opposite lateral edges, respectively, of the upper portion 114D of the dielectric 114 as shown and connected to voltage source 116E which applies a voltage between them to induce the desired asymmetry in the structure. Alternatively, the electrodes 112D, 112E could be placed laterally along the bottom portion 114E of the dielectric 114, the distinct portions of the dielectric material still providing the asymmetry being above and below the strip.

Figure 46 shows an example of a magneto-optic modulator wherein the waveguiding strip 110 and overlying electrode 112F are used to carry a current  $I$  in the opposite directions shown. The dielectric material surrounding the metal waveguide strip 110 exhibits a magneto-optic effect or is a ferrite. The magnetic fields generated by the current  $I$  add in the dielectric portion between the electrodes 110 and 112F and essentially cancel in the portions above the top electrode 112F and below the waveguide 110. The applied magnetic field thus induces the desired asymmetry in the waveguiding structure. The electrode 112F is placed far enough from the guiding strip 110 that optical coupling between the strips is negligible.

Figure 47 depicts a thermo-optic modulator wherein the waveguiding strip 110 and the overlying electrode 112G are maintained at temperatures  $T_2$  and  $T_1$ , respectively. The dielectric material 114 surrounding the metal waveguide exhibits a thermo-optic effect. The temperature difference creates a thermal gradient in the dielectric portion 114G between the electrode 112G and the strip 110. The variation in the applied temperature thus induce the desired asymmetry in the waveguiding structure. The electrode 112G is placed far enough from the guiding strip 110 that optical coupling between the strips is negligible.

It should be appreciated that the modulator devices described above with reference to Figures 43(a) to 47 may also serve as variable optical attenuators with the attenuation being controlled via the external stimulus, i.e. voltage, current, temperature, which varies the electromagnetic property.

Figures 48, 49 and 50 depict optical switches that operate on the principle of “split and attenuate”. In each case, the input optical signal is first split into N outputs using a power divider; a one-to-two power split being shown in Figures 48, 49 and 50. The undesired outputs are then “switched off” or highly attenuated by inducing a large asymmetry in the corresponding output waveguides. The asymmetry must be large

enough to completely cut-off the main purely bound long-ranging mode supported by the waveguides. The asymmetry is induced by means of overlying electrodes as in the waveguide structures of Figures 43, 46 or 47, respectively.

In the switches shown in Figures 48, 49 and 50, the basic waveguide configuration is the same and comprises an input waveguide section 120 coupled to two parallel branch sections 122A and 122B by a wedge-shaped splitter 124. All four sections 120, 122A, 122B and 124 are co-planar and embedded in dielectric material 126. The thickness of the metal film is  $d_3$ . Two rectangular electrodes 128A and 128B, each of thickness  $d_1$ , are disposed above branch sections 122A and 122B, respectively, and spaced from them by a thickness  $d_2$  of the dielectric material 126 at a distance large enough that optical coupling between the strips is negligible. Each of the electrodes 128A and 128B is wider and shorter than the underlying metal film 122A or 122B, respectively. In the switch shown in Figure 48, the asymmetry is induced electro-optically by means of a first voltage source 130A connected between metal film 122A and electrode 128A for applying voltage  $V_1$  therebetween, and a second voltage source 130B connected between metal film 122B and electrode 128B, for applying a second voltage  $V_2$  therebetween. In the switch shown in Figure 49, the asymmetry is induced magneto-optically by a first current source 132A connected between metal film 122A and electrode 128A, which are connected together by connector 134A to complete the circuit, and a second current source 132B connected between metal film 122B and electrode 128B, which are connected together by connector 134B to complete that circuit.

In the switch shown in Figure 50, the asymmetry is induced thermo-optically by maintaining the metal strips 122A and 122B at temperature  $T_0$  and the overlying electrodes 128A and 128B at temperatures  $T_1$  and  $T_2$ , respectively.

It will be appreciated that, in the structures shown in Figures 48, 49 and 50, the dielectric surrounding the metal strip will be electro-optic, magneto-optic, or thermo-optic, or a magnetic material such as a ferrite, as appropriate.

In general, any of the sources, whether of voltage, current or temperature, may be variable.

Although the switches shown in Figures 48, 49 and 50 are 1 x 2 switches, the invention embraces 1 x N switches which can be created by adding more branch sections and associated electrodes, etc.

It will be appreciated that, where the surrounding material is acousto-optic, the external stimulus used to induce or enhance the asymmetry could be determined by analogy. For example, a structure similar to that shown in Figure 45 could be used with the electro-optic material replaced by acousto-optic material and the electrodes 112D and 112E used to apply compression or tension to the upper portion 114D.

To facilitate description, the various devices embodying the invention have been shown and described as comprising several separate sections of the novel waveguide structure. While it would be feasible to construct devices in this way, in practice, the devices are likely to comprise continuous strips of metal or other high charge carrier density material, i.e. integral strip sections, fabricated on the same substrate.

The foregoing examples are not meant to be an exhaustive listing of all that is possible but rather to demonstrate the breadth of application of the invention. The inventive concept can be applied to various other elements suitable for integrated optics devices. It is also envisaged that waveguide structures embodying the invention could 10 be applied to multiplexers and demultiplexers.

#### INDUSTRIAL APPLICABILITY

Embodiments of the invention may be useful for signal transmission and routing or to construct components such as couplers, power splitters/combiners, interferometers, 15 modulators, switches, periodic structures and other typical components of integrated optics.

Although embodiments of the invention have been described and illustrated in detail, it is to be clearly understood that the same is by way of illustration and example only and not to be taken by way of the limitation, the spirit and scope of the present 20 invention being limited only by the appended claims.

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## CLAIMS

1. An optical device characterized by a waveguide structure formed by a thin strip (100) of a material having a relatively high free charge carrier density surrounded by a material having a relatively low free charge carrier density, the strip having finite width (W) and thickness (t) with dimensions such that optical radiation having a wavelength in a predetermined range couples to the strip and propagates along the length of the strip as a plasmon-polariton wave.
- 10 2. An optical device according to claim 1, characterized in that said free charge carrier density of the surrounding material is substantially negligible.
3. A device according to claim 1, for optical radiation having a free-space wavelength near 1550 nm, characterized in that the strip (100) comprises a metal and has thickness less than about 0.1 microns and width of a few microns.
- 15 4. A device according to claim 3, characterized in that the strip (100) has a thickness of about 20nm and width of about 4 microns.
- 20 5. A device according to claim 1, characterized in that the strip (100;104;106;108;130) is straight, curved, bent, or tapered.
6. A device according to claim 1, further characterized by at least one additional said waveguide structure comprising a second said strip (108), wherein the second strip 25 (108) is curved and the first-mentioned strip (100) is offset outwardly (O<sub>1</sub>,O<sub>2</sub>) relative to an axis of curvature of the second strip (108).
7. A device according to claim 6, characterized in that the first (100) and second (108) strips are separate and juxtaposed with one end of the first strip adjacent an end 30 of the second strip that is offset outwardly (O<sub>1</sub>,O<sub>2</sub>) relative to the end of the first strip.
8. A device according to claim 6, characterized in that said first (100) and second (108) strips are integral with each other.
- 35 9. A device according to claim 1, further characterized by a plurality of branch waveguide structures of similar construction to the first waveguide structure, wherein the plurality of branch waveguide structures comprise a plurality of branch strips (104), respectively, having ends coupled to an end portion of the first-mentioned strip (100;106)

thereby forming a combiner/splitter (110), the arrangement being such that said optical radiation leaving said first-mentioned strip via said one end portion will be split between said plurality of branch strips and conversely optical radiation coupled to said one end portion by said plurality of branch strips will be combined to leave said first-mentioned 5 strip by an opposite end.

10. A device according to claim 9, characterized in that the branch strips (104) are integral with the first strip (100;106).

10 11. A device according to claim 9, characterized in that branch strips (104) are distinct from the first strip (100;106).

12. A device as according to claim 1, further characterized by a first plurality of branch waveguide structures each similar in construction to the first waveguide structure  
15 and having a corresponding plurality of branch strips forming a splitter (110;116), and a second plurality of waveguide structures similar to the first waveguide structure and comprising a second plurality of branch waveguide structures similar in construction to the first waveguide structure, and comprising a corresponding second plurality of branch strips, and an output waveguide structure similar in construction to the first waveguide  
20 structure and having an output strip coupled to respective ends of the second plurality of branch strips to form a combiner (110;116), the splitter and combiner having their respective branch strips connected together to form a Mach-Zehnder interferometer (118;120), the arrangement being such that optical radiation input via said input strip produces a plasmon-polariton wave portion propagating along each of said first plurality  
25 of branch strips sections, the plasmon- polariton waves being coupled to the second plurality of branch strips and recombined to propagate along said output strip.

13. A device according to claim 12, further characterized by means (122,124,126) for adjusting the propagation characteristics of at least one of said plasmon-polariton  
30 waves as it propagates along the branch strip.

14. A device according to claim 13, characterized in that the adjusting means (122,124,126) is arranged to modulate said propagation characteristics so as to obtain destructive interference upon recombination and thereby modulate the intensity of said  
35 optical radiation.

15. A device according to claim 14, characterized in that the adjusting means comprises at least one electrode (122,124) adjacent one of said branch strips and a

voltage source (126) for providing a potential difference between the electrode and the branch strip.

16. A device according to claim 14, characterized in that the material around the strip  
5 is an electro-optic material and the adjusting means comprises a pair of electrodes (122,124) one each side of the strip and a voltage source (126) connected to the electrodes for providing a potential difference therebetween.

17. A device according to claim 14, characterized in that the adjusting means is  
10 arranged to induce a magnetic field in the strip.

18. A device according to claim 17, characterized in that the adjusting means comprises a coil formed by metal-plated via holes and surface conductors.

15 19. A device according to claim 17, characterized in that the adjusting means comprises a solenoid having magnetic poles either side of the strip.

20. A device according to claim 1, characterized by a plurality of said waveguide structures comprising a corresponding plurality of strips and arranged to form a periodic  
20 structure comprising a plurality of unit cells (134), each unit cell comprising a first strip (100) and a second strip (100'), the first and second strips being dissimilar in one or more of shape and size.

21. A device according to claim 20, characterized in that the plurality of strips  
25 (100,100') are integral with each other.

22. A device according to claim 20, characterized in that the strips (100,100') are separated from each other by a predetermined distance.

30 23. A device according to claim 20, further characterized by adjusting means for modifying an optical transfer function of the device.

24. A device according to claim 23, characterized in that the adjusting means comprises a voltage source for providing a potential difference between the first and  
35 second strips in each unit cell.

25. A device according to claim 24, characterized in that the material between the strips is an electro-optic material and the voltage source provides a potential gradient therein.

5 26. A device according to claim 24, characterized in that the adjusting means comprises at least one electrode positioned near the periodic structure and connected to one terminal of the voltage source, a second terminal of the voltage source being connected to at least one of the first and second strips of each unit cell.

10 27. A device according to claim 1, characterized by a plurality of said waveguide structures, their strips arranged to form an edge coupler, two (100") of said strips being disposed with their adjacent edges in close proximity such that at least some of said plasmon-polariton wave propagating along one of said strips will couple onto the other of the strips.

15 28. An edge coupler according to claim 27, further characterized by means (126) for adjusting the propagation characteristics of said plasmon-polariton wave propagating along the coupled strips so as to control the degree of coupling between the strips.

20 29. An edge coupler according to claim 27, characterized in that the material between the coupled strips is electro-optic and further comprising means (126) for adjusting the charge on the strips and the refractive index of the material therebetween.

30. An edge coupler according to claim 27, characterized in that the first and second  
25 strips are not coplanar.

31. An edge coupler according to claim 27, characterized in that the material between the coupled strips is electro-optic and further comprising means for adjusting the charge on the strips and the refractive index of the material therebetween.

30 32. A device according to claim 1, characterized by at least three of said waveguide structures (104) arranged to form a waveguide intersection (142), their respective strips each having one end connected to the other strips to form said intersection, distal ends of the three strips constituting ports such that optical radiation input via the distal end of  
35 one of the strips will be conveyed across the intersection to emerge from one or both of the other strips.

33. A device according to claim 1, characterized in that the surrounding material is inhomogeneous.

34. A device according to claim 33, characterized in that the surrounding material 5 comprises a combination of slabs, strips, laminae, or continuously variable material composition.

35. A device according to claim 33, characterized in that the strip is inhomogeneous.

10 36. A device according to claim 34, characterized in that the strip is inhomogeneous.

37. A device according to claim 1, characterized in that the strip is inhomogeneous.

38. An optical device characterized by a waveguide structure formed by a thin strip 15 of a material having a relatively high free charge carrier density surrounded by a material having a relatively low free carrier density, the strip having finite width and thickness with dimensions such that optical radiation having a wavelength in a predetermined range couples to the strip and propagates along the length of the strip as a plasmon-polariton wave, wherein the material comprises two distinct portions with the 20 strip extending therebetween, at least one of the two distinct portions having at least one variable electromagnetic property, and that the device further comprises means for varying the value of said electromagnetic property of said one of the portions so as to vary the propagation characteristics of the waveguide structure and the propagation of the plasmon-polariton wave.

25

39. A device according to claim 38, characterized in that, for one said value of the electromagnetic property for said one of the portions propagation of the plasmon-polariton wave is supported and for another value of said electromagnetic property of said one of said portions propagation of the plasmon-polariton wave is at least inhibited.

30

40. A device according to claim 38, characterized in that said means for varying the electromagnetic property changes the size of at least one of said portions.

41. A device according to claim 38, characterized in that one of said portions is a 35 fluid.

42. A device according to claim 38, characterized in that said electromagnetic property is permittivity and the varying means (116;116A;116B;116C;116E) varies the

permittivity by inducing a change in one or more of an electrical field in material of said portion, magnetic field in material of said portion, mechanical strain in material of said portion, and temperature in the material (114D) of said portion.

5 43. A device according to claim 38, characterized in that said electromagnetic property is permeability and the varying means varies the permeability by inducing a change in one or more of a magnetic field in material of said portion; mechanical strain in the material of said portion, and temperature in the material of said portion.

10 44. An optical device according to claim 38, characterized in that said free charge carrier density of the surrounding material is substantially negligible.

45. A device according to claim 38, for optical radiation having a free-space wavelength near 1550 nm, characterized in that the strip comprises a metal and has  
15 thickness less than about 0.1 microns and width of a few microns.

46. A device according to claim 38, characterized in that the strip is straight, curved, bent, or tapered.

20 47. A device according to claim 38, characterized in that the material is electro-optic and the varying means comprises an electrode (112) overlying said one of said portions and means for applying a potential difference between the electrode and the strip.

48. A device according to claim 38, characterized in that the material is electro-optic  
25 and the varying means comprises first and second electrodes disposed one at each side of the strip, said one of the portions being between the first electrode and the strip and the other of said portions being between the second electrode and the strip, and means for applying a potential difference between the strip and at least one of the first and second electrodes.

30

49. A device according to claim 48, characterized in that the applying means comprises a first voltage source for applying a first potential difference between the strip and the first electrode and a second voltage source for applying a second potential difference between the strip and the second electrode.

35

50. A device according to claim 48, characterized in that the applying means comprises means for coupling one terminal of a voltage source to the strip and a second terminal of the voltage source in common to the first and second electrodes.

51. A device according to claim 38, characterized in that the material is electro-optic, the strip is embedded in the material with the said one of the portions adjacent one surface of the strip, and the varying means comprises first and second electrodes disposed laterally of the strip at opposite sides of said one of said portions and means of applying a potential difference between the electrodes, the other of said portions being adjacent an opposite surface of the strip.

52. A device according to claim 38, characterized in that the material is magneto-optic and the varying means comprises means for establishing a current flowing in at least one of the strip and an adjacent electrode, the said one of the portions being between the electrode and the strip.

53. A device according to claim 38, characterized in that the material is thermo-optic, at least one electrode is provided adjacent to the strip with said one of the portions therebetween, and the varying means comprises means for establishing a temperature difference between the strip and the electrode.

54. A device according to claim 38, further characterized by a plurality of waveguide structures similar in construction to the first-mentioned structure and each comprising one of a plurality of said strips, the plurality of strips having respective proximal ends juxtaposed to one end of the first-mentioned strip to form a combiner/splitter, the arrangement being such that said optical radiation leaving said first-mentioned strip via said one end will be split between said plurality of strips and conversely said optical radiation coupled to said one end by said plurality of strips will be combined to leave said first-mentioned strip by an opposite end, wherein the varying means is coupled to at least one of the plurality of strips.

55. A device according to claim 54, characterized in that the material is electro-optic and the waveguide structures comprise an input strip for receiving said optical radiation at one end thereof and end-coupled to a splitter at an opposite end thereof, first and second branch strips each having a proximal end coupled to the splitter for receiving a portion of the radiation, the varying means comprising an electrode adjacent a respective one of the branch strips with said one of the portions therebetween and means for applying a potential difference between the electrode and said one of the branch strips.

35

56. A device according to claim 55, characterized in that the varying means further comprises a second electrode adjacent the other branch strip with a second one of said

portions therebetween and means for applying a second potential difference between the second electrode and the second branch strip.

57. A device according to claim 54, characterized in that the material is magneto-optic and the waveguide structures comprise an input strip for receiving said optical radiation at one end thereof and end-coupled to a splitter at an opposite end thereof, first and second branch strips each having a proximal end coupled to the splitter for receiving a portion of the radiation, the varying means comprising an electrode adjacent a respective one of the branch strips with said one of the portions therebetween and means 10 for establishing a current flowing in said electrode and said one of the branch strips.

58. A device according to claim 57, characterized in that the varying means further comprises a second electrode adjacent the other branch strip with a second one of said portions therebetween and means for establishing a second current flowing in the second 15 electrode and the second branch strip.

59. A device according to claim 54, characterized in that the material is thermo-optic and the waveguide structures comprise an input strip for receiving said optical radiation at one end thereof and end-coupled to a splitter at an opposite end thereof, first and 20 second branch strips each having a proximal end coupled to the splitter for receiving a portion of the radiation, the varying means comprising an electrode adjacent a respective one of the branch strips with said one of the portions therebetween and means for establishing a temperature difference between said electrode and said one of the branch strips.

25

60. A device according to claim 59, characterized in that the varying means further comprises a second electrode adjacent the other branch strip with a second one of said portions therebetween and means for establishing a second temperature difference between the second electrode and the second branch strip.

30

61. A device according to claim 38, characterized in that the varying means comprises a coil formed by metal-plated via holes and surface conductors.

62. A device according to claim 38, characterized in that the varying means 35 comprises a solenoid having magnetic poles either side of the strip.

63. A device according to claim 20, comprising an array of said periodic structures disposed adjacent each other and having strips sized and shaped so as to form stop bands at desired locations in the optical spectrum.

5 64. A device according to claim 63, characterized in that the array is two-dimensional.

65. A device according to claim 63, characterized in that the array is three-dimensional.

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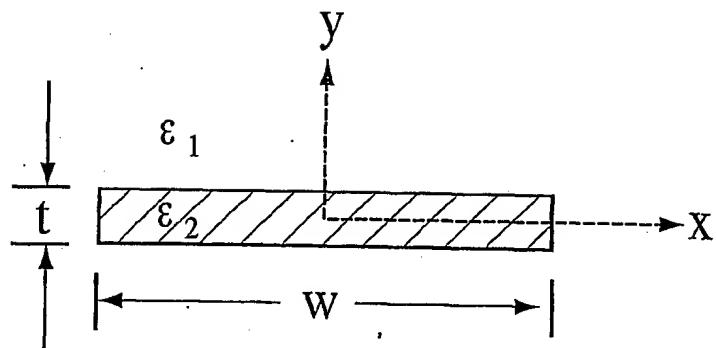


FIG. 1 (a)

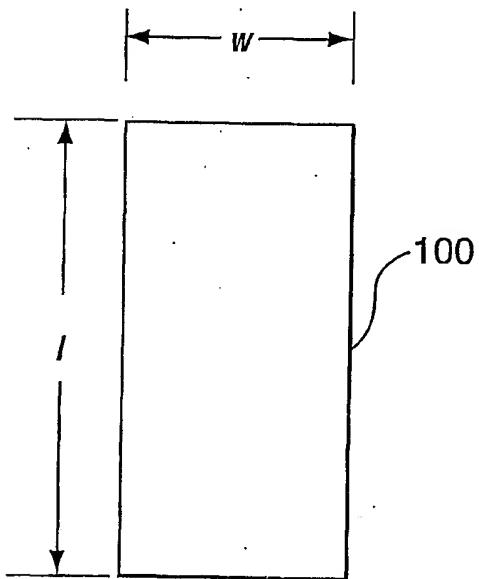


FIG. 1 (b)

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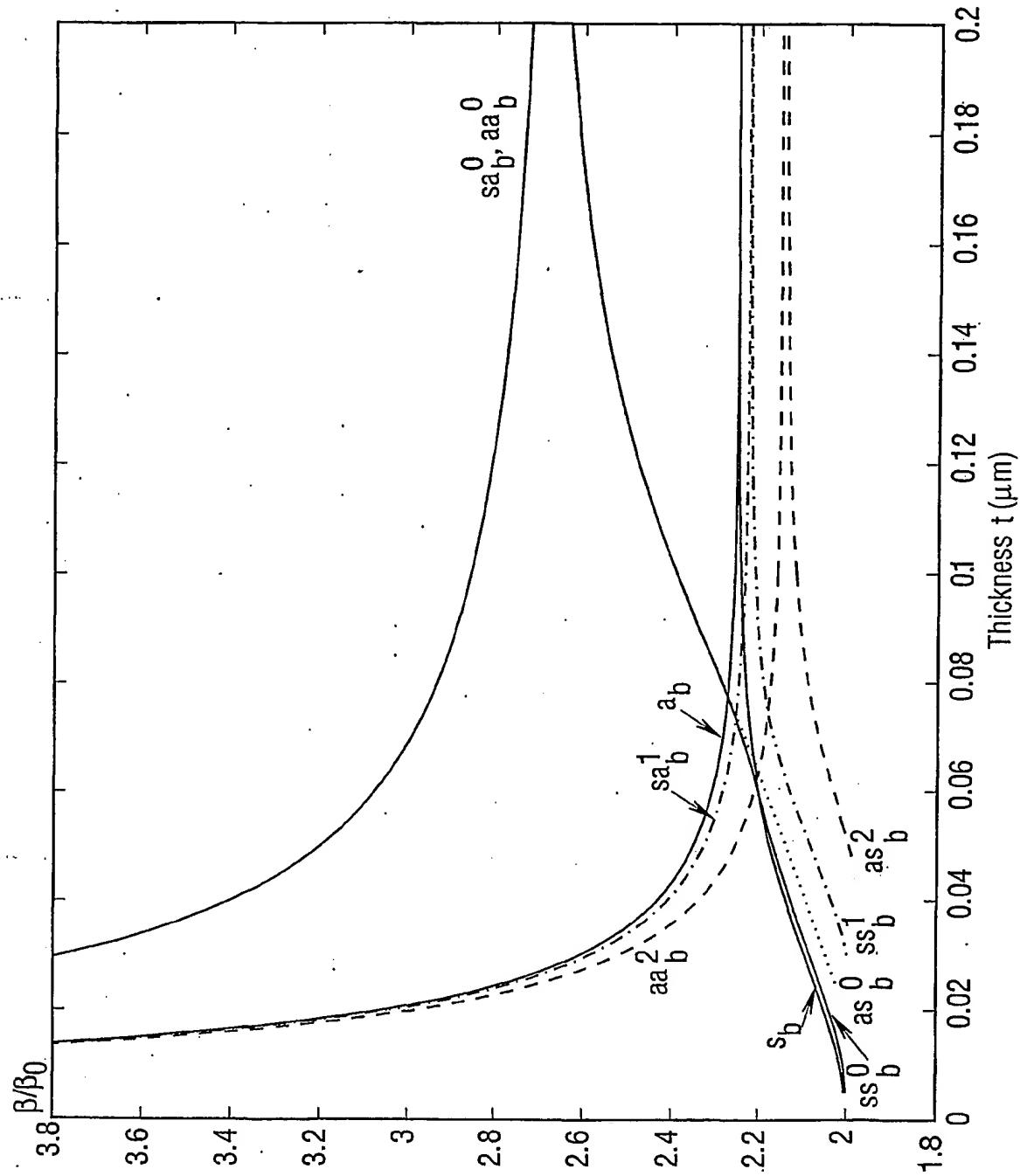


FIG. 2 (a)

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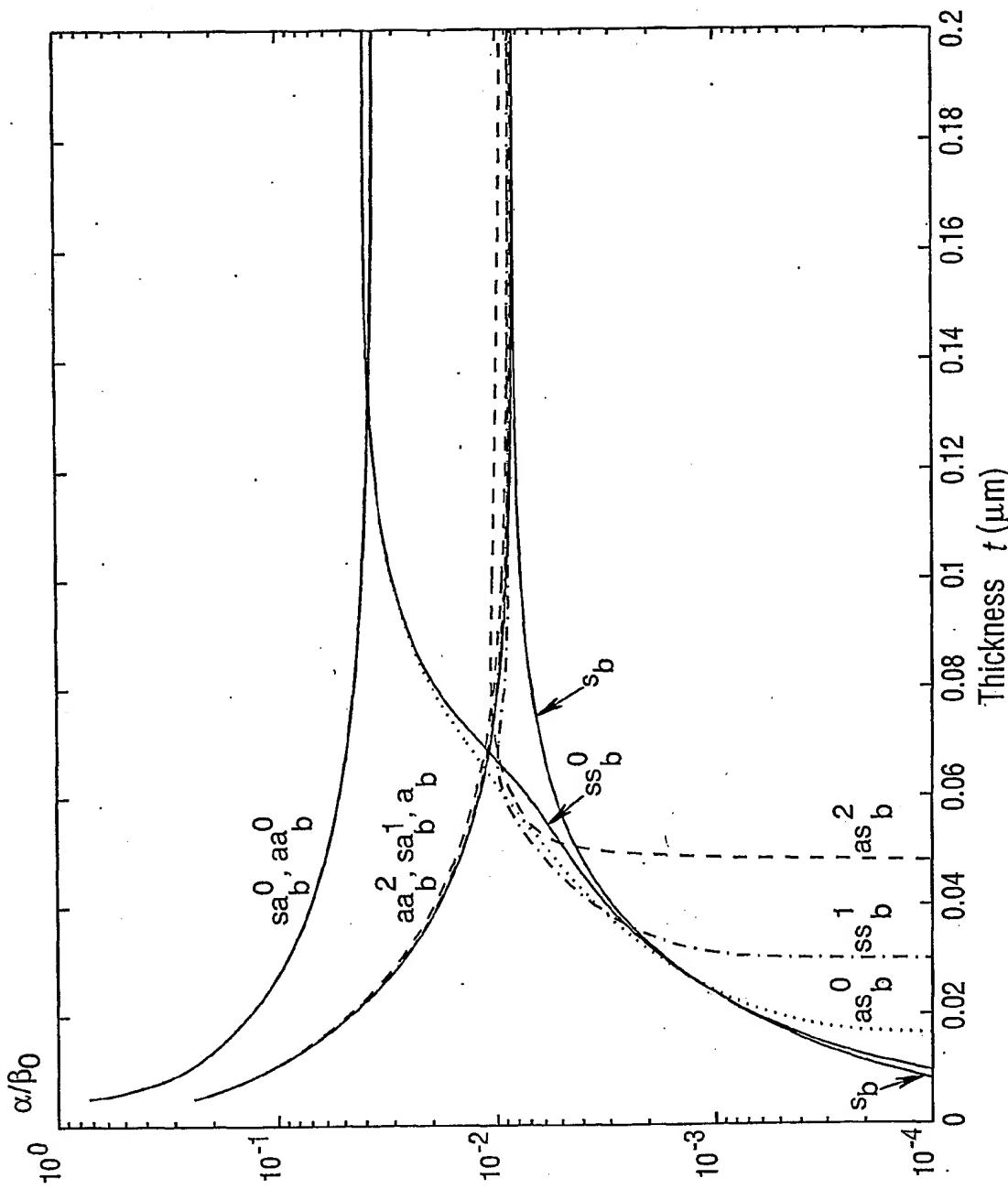
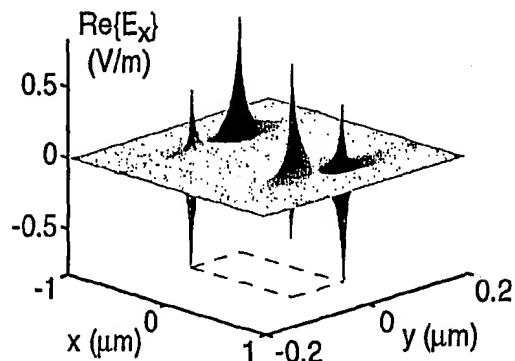
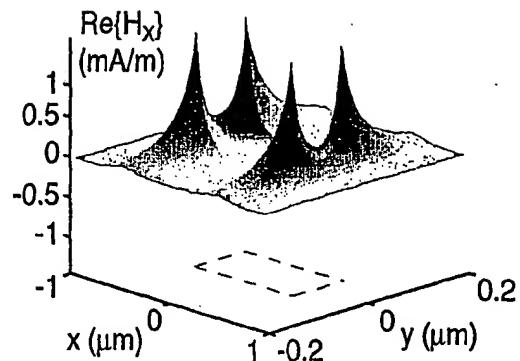
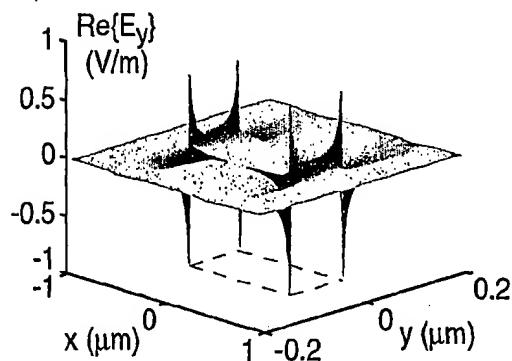
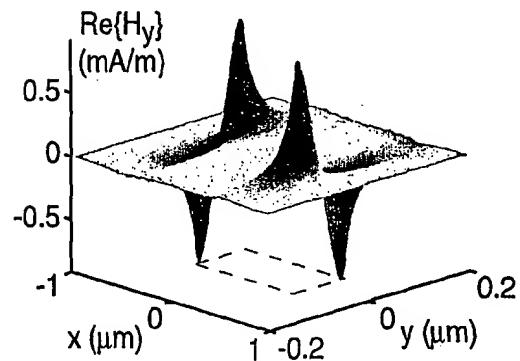
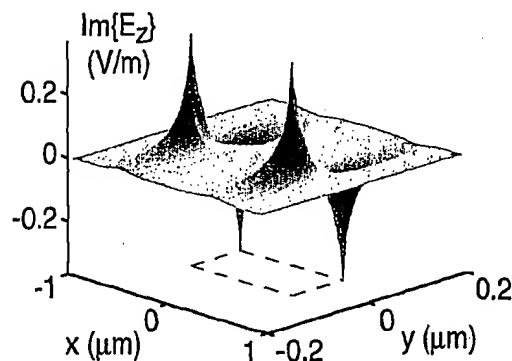
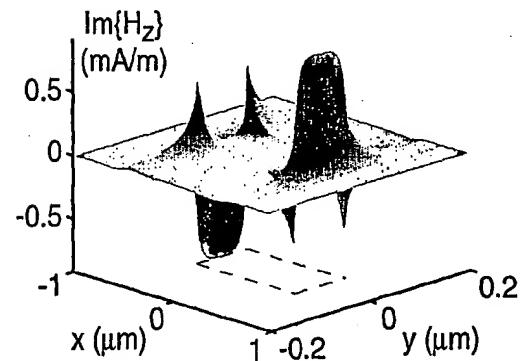
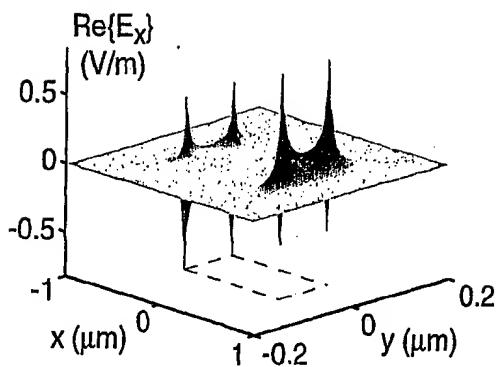
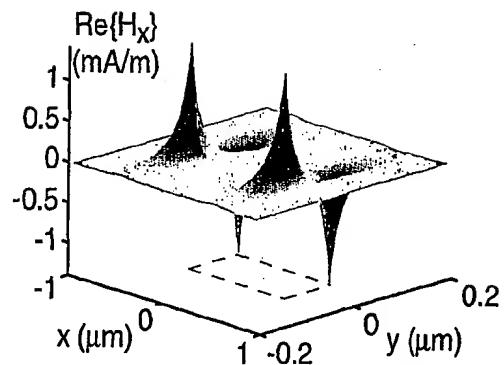
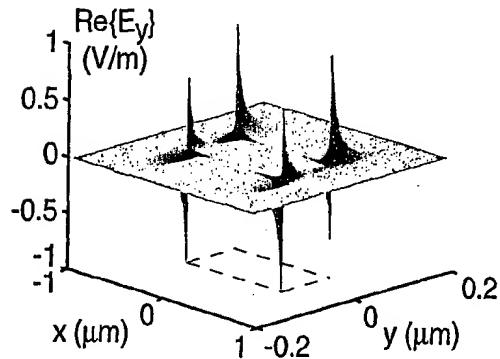
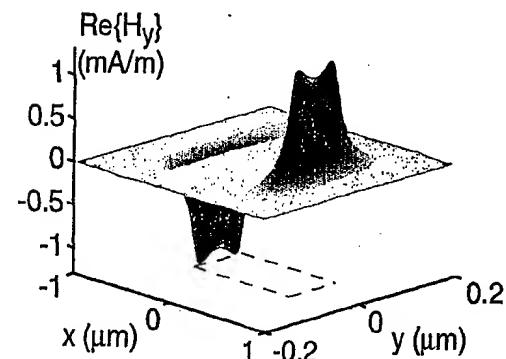
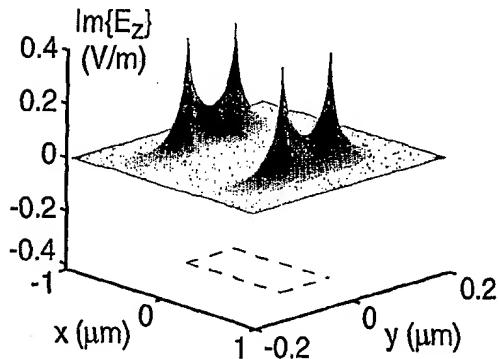
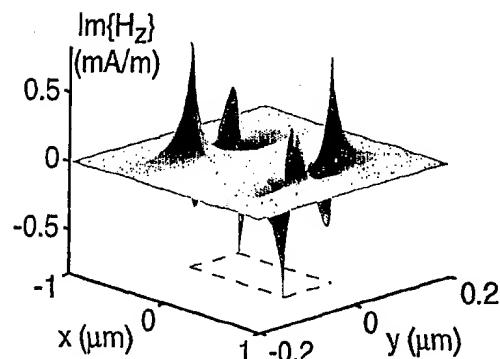


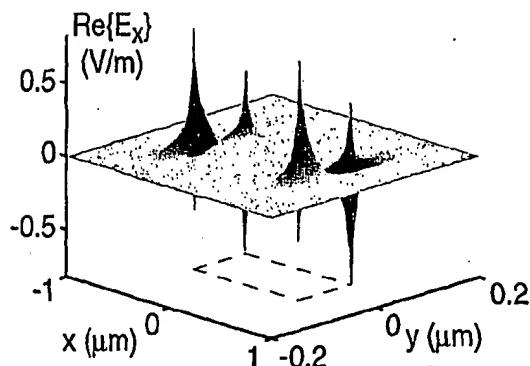
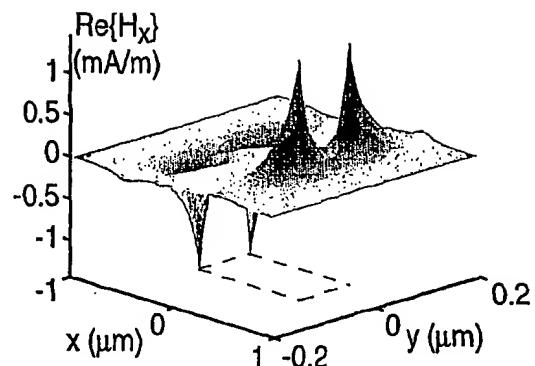
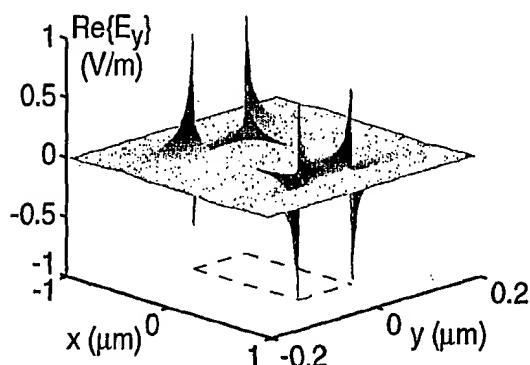
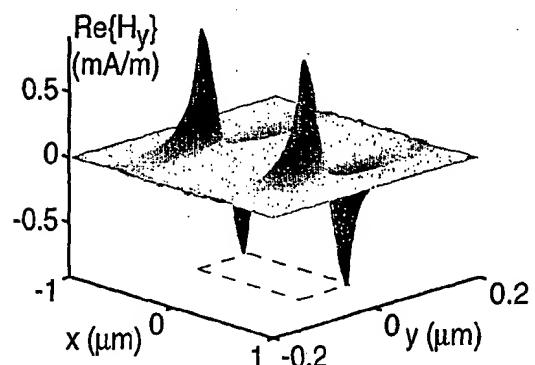
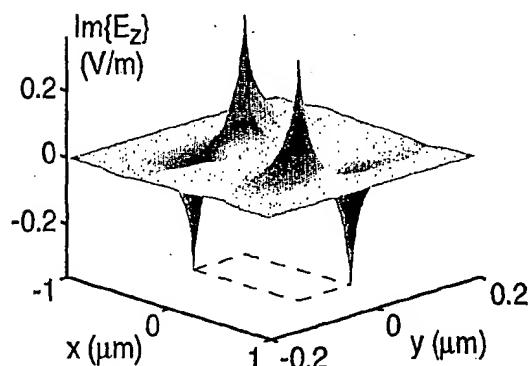
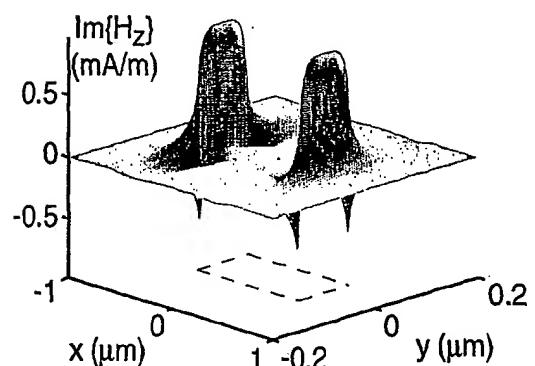
FIG. 2(b)

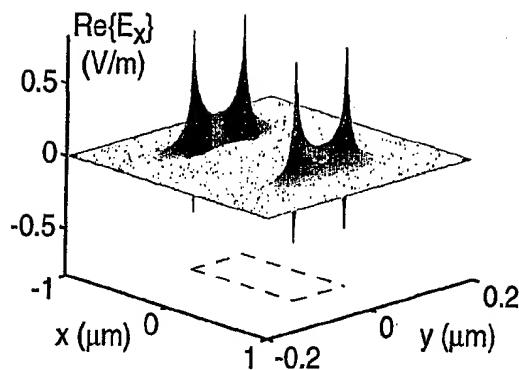
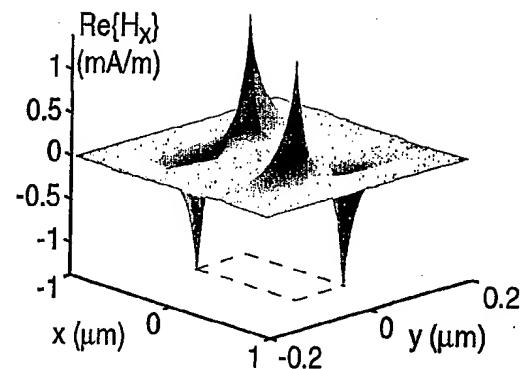
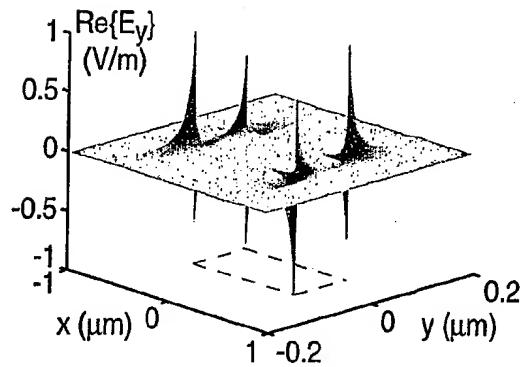
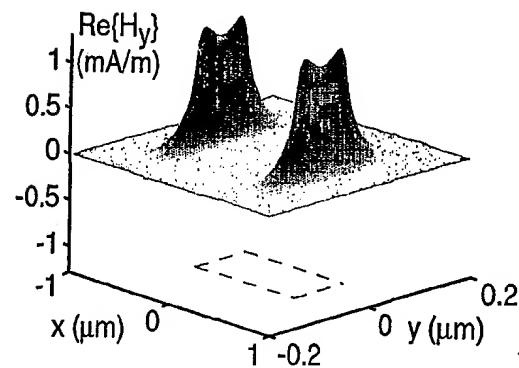
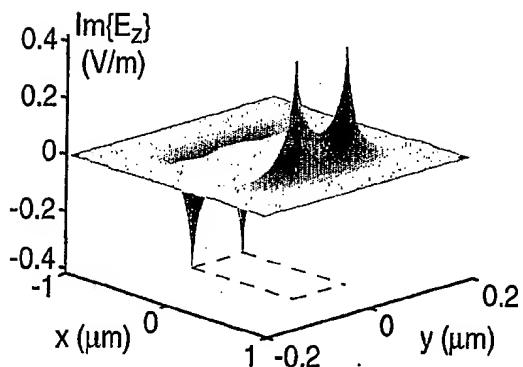
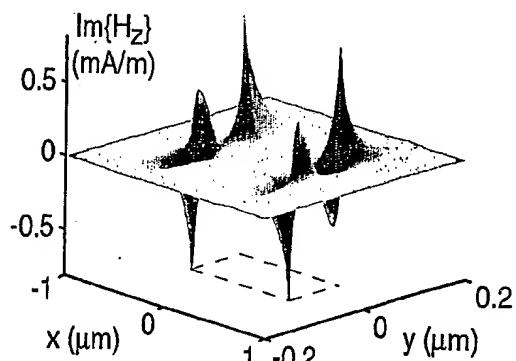
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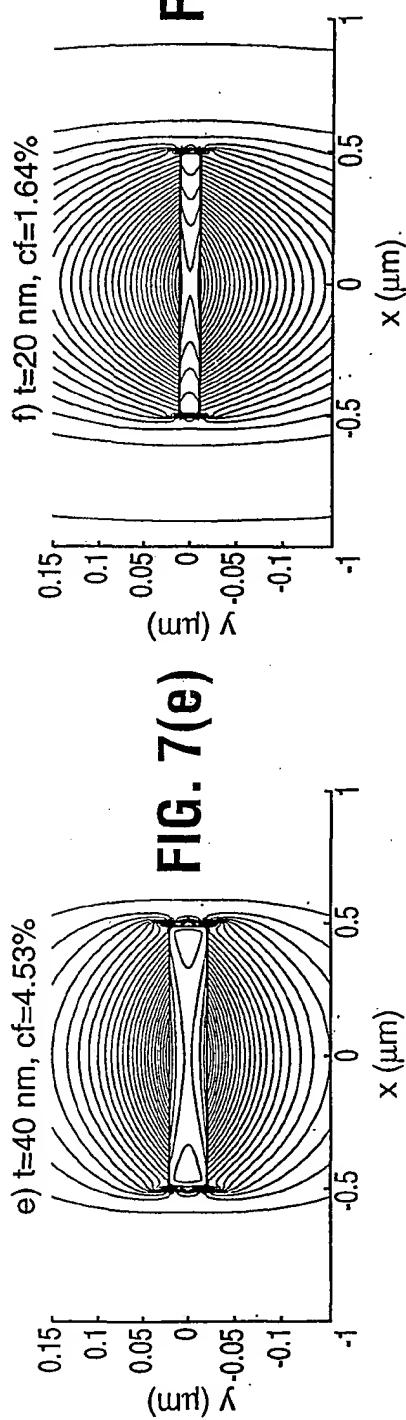
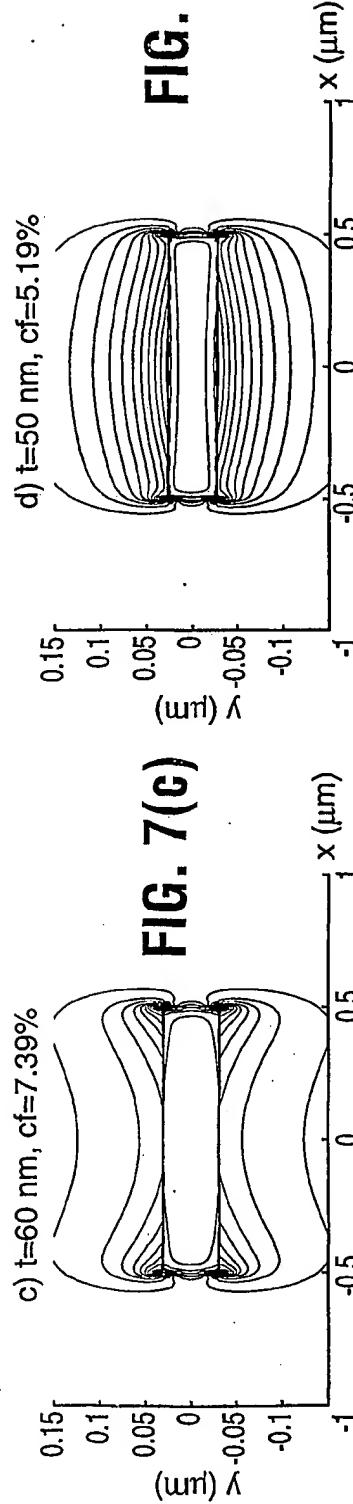
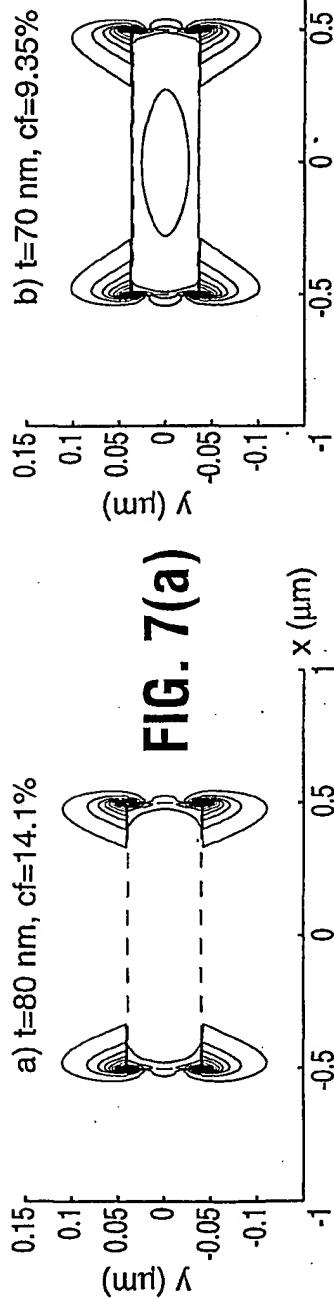
**FIG. 3(a)****FIG. 3(b)****FIG. 3(c)****FIG. 3(d)****FIG. 3(e)****FIG. 3(f)**

**FIG. 4(a)****FIG. 4(b)****FIG. 4(c)****FIG. 4(d)****FIG. 4(e)****FIG. 4(f)**

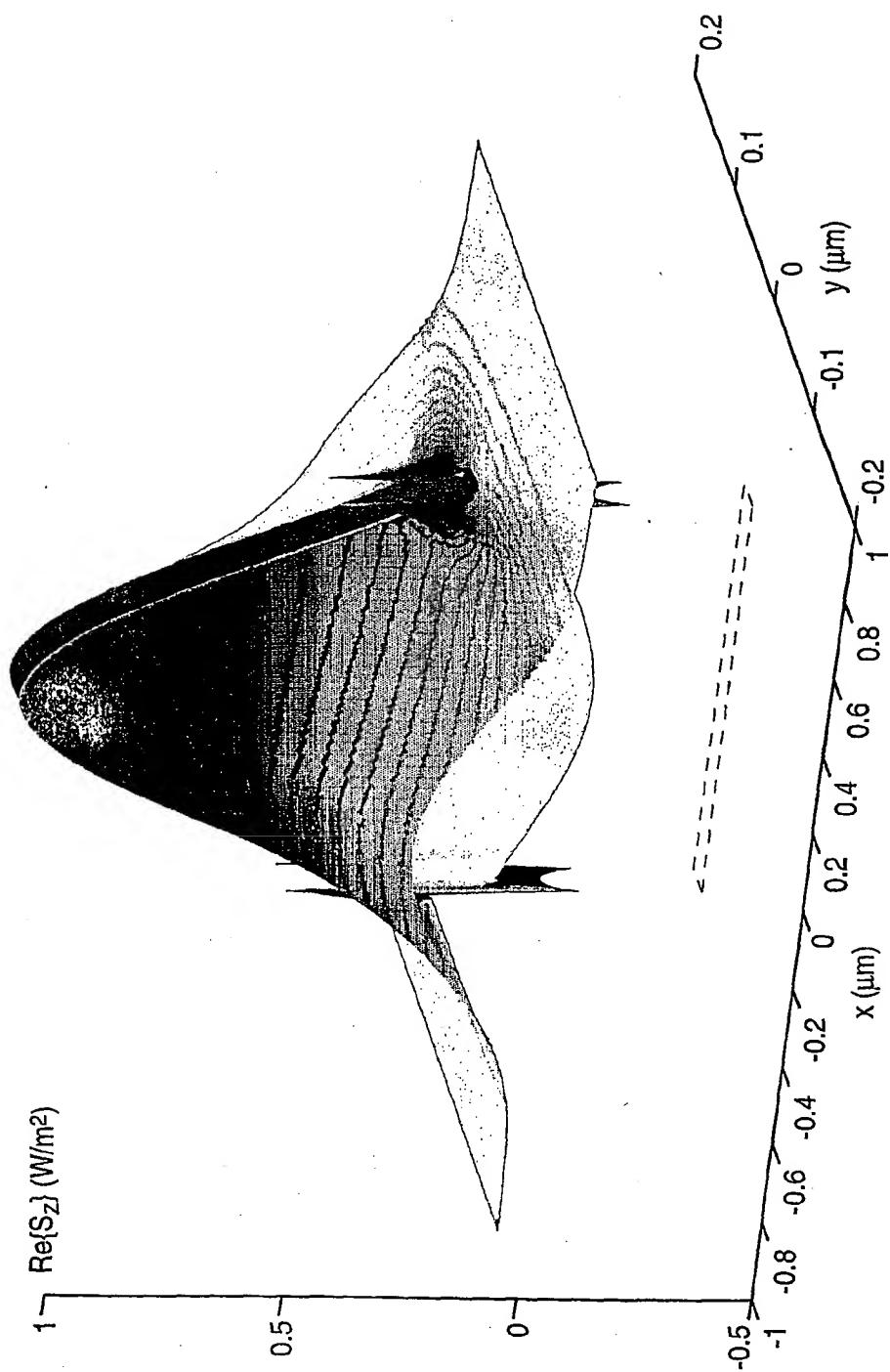
**FIG. 5(a)****FIG. 5(b)****FIG. 5(c)****FIG. 5(d)****FIG. 5(e)****FIG. 5(f)**

**FIG. 6(a)****FIG. 6(b)****FIG. 6(c)****FIG. 6(d)****FIG. 6(e)****FIG. 6(f)**

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**FIG. 8****SUBSTITUTE SHEET (RULE 26)**